



Efficient encoder design of LDPC code using circulant matrix and eIRA codes

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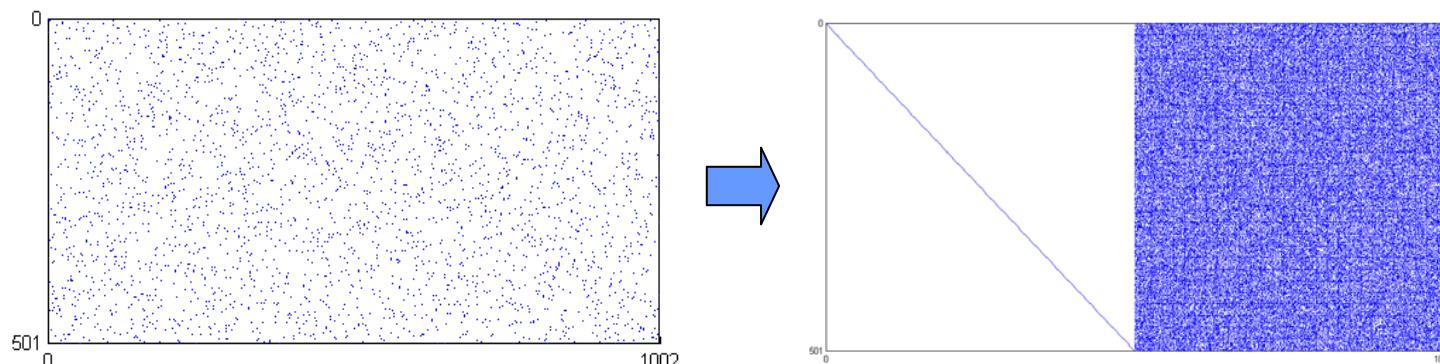


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Introduction

□ Encoding of LDPC codes

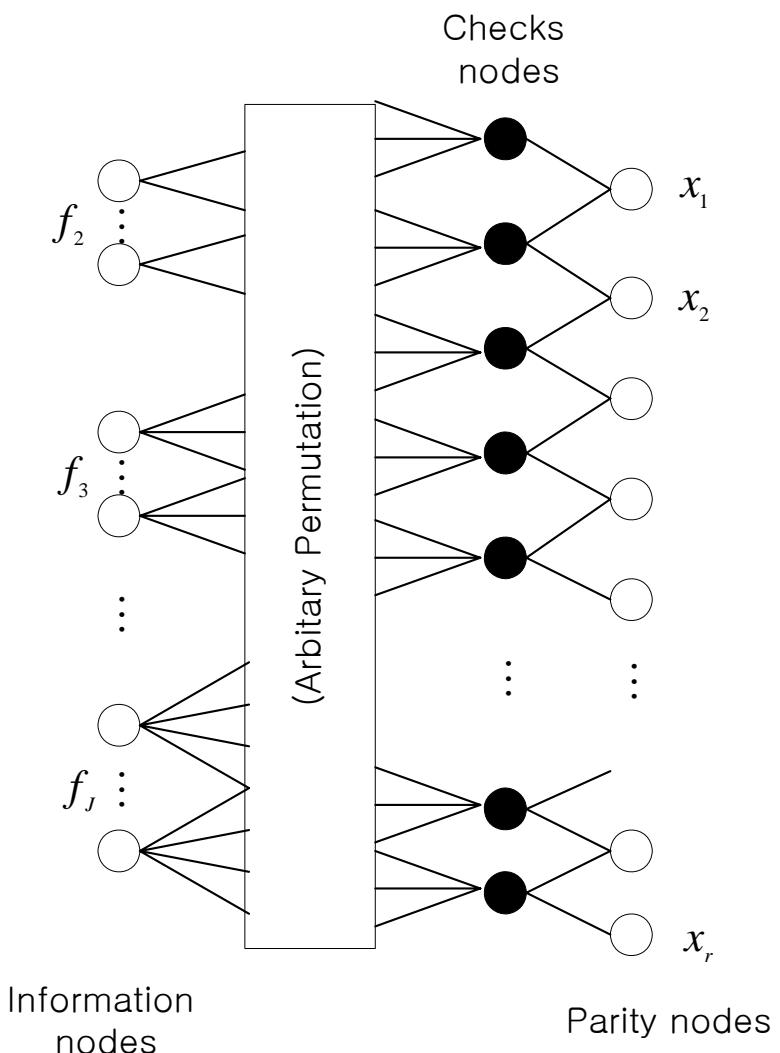
- Information bit & Parity bit are computed by matrix multiplication in linear block code.
- 1's in generator matrix are more than parity-check matrix
- Encoding complexity is much high with matrix multiplication operation.



Design of Structural Codes(1/2)

□ Irregular Repeated-Accumulate (IRA) codes

- k : number of variable nodes
- r : number of checks nodes
- r : number of parity nodes
- Each check node is connected to a information nodes
- The value of parity bit is determined uniquely by the condition that the mod-2 sum of the values of the variable nodes connected to each of the check nodes is zero.





Design of Structural Codes(2/2)



□ Extended Irregular Repeated-Accumulate (eIRA) codes

- Reference : Michael Yang, “Design of Efficiently Encodable Moderate-Length High-Rate Irregular LDPC Codes”, *IEEE Trans. Comm.* Vol. 52. pp 564-571, April. 2004
- Complexity is reduced using differential encoder structure of IRA codes
- $\mathbf{H} = [\mathbf{H}_1 \mid \mathbf{H}_2]$
 - \mathbf{H}_1 : $(n-k) \times k$ sparse matrix (using density evolution)
 - \mathbf{H}_2 : $(n-k) \times (n-k)$ dual diagonal matrix

$$\mathbf{H}_2 = \begin{pmatrix} 1 & & & & \\ 1 & 1 & & & \\ & 1 & 1 & & \\ & & \ddots & \ddots & \\ & & & 1 & 1 \\ & & & & 1 & 1 \end{pmatrix} \quad \mathbf{H}_2^{-T} = \begin{pmatrix} 1 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 1 & 1 \\ \dots & \dots & \dots & \dots & \dots \\ & & 1 & 1 & \\ & & & & 1 \end{pmatrix}$$



Modified Encoder Structure(1/6)



□ Circulant matrix

- H_1 matrix is constructed by $p \times q$ matrix blocks which are $l \times l$ circulant submatrix , where p is a number of row blocks, q is a number of column blocks.
- Each weight in each circulant submatrix can be different.

$$H_1 = \begin{pmatrix} h_{11} & h_{12} & \cdots & h_{1q} \\ h_{21} & h_{22} & & h_{2q} \\ \vdots & & \ddots & \vdots \\ h_{p1} & h_{p2} & \cdots & h_{pq} \end{pmatrix}$$

- Proposed H_1 matrix can be more modified by row permutation matrix.
- Row permutation matrix P is $(n - k) \times (n - k)$ matrix and each rows and columns has only one 1's.
- To prevent from performance degradation, modify to $H_1' = PH_1$ using row permutation matrix.



Modified Encoder Structure(2/6)



□ Parity-check matrix

$$H = [H_1 | H_2]$$
$$= [PH_1 | H_2]$$

$$= \left[\begin{array}{cccccc|c} & & & h_{11} & h_{12} & \cdots & \cdots & h_{1q} & 1 \\ & 1 & & h_{21} & h_{22} & \cdots & \cdots & h_{2q} & 1 & 1 \\ & \ddots & & 1 & \times & \vdots & \vdots & \ddots & \ddots & \ddots & \vdots \\ & 1 & & 1 & & \ddots & \ddots & \ddots & & \ddots & \ddots \\ & & & h_{p1} & h_{p2} & \cdots & \cdots & h_{pq} & & & 1 & 1 \\ \end{array} \right]$$

Modified Encoder Structure(3/6)

□ Generator matrix

- $\mathbf{H} = [\mathbf{H}_1' \mid \mathbf{H}_2] = [\mathbf{H}_2^{-1}\mathbf{H}_1' \mid \mathbf{I}_{n-k}] \rightarrow \mathbf{G} = [\mathbf{I}_k \mid \mathbf{H}_1'^T \mathbf{H}_2^{-T}]$

$$\mathbf{G} = [I \mid \mathbf{H}_1^T \mathbf{P}^T \mathbf{H}_2^{-T}]$$

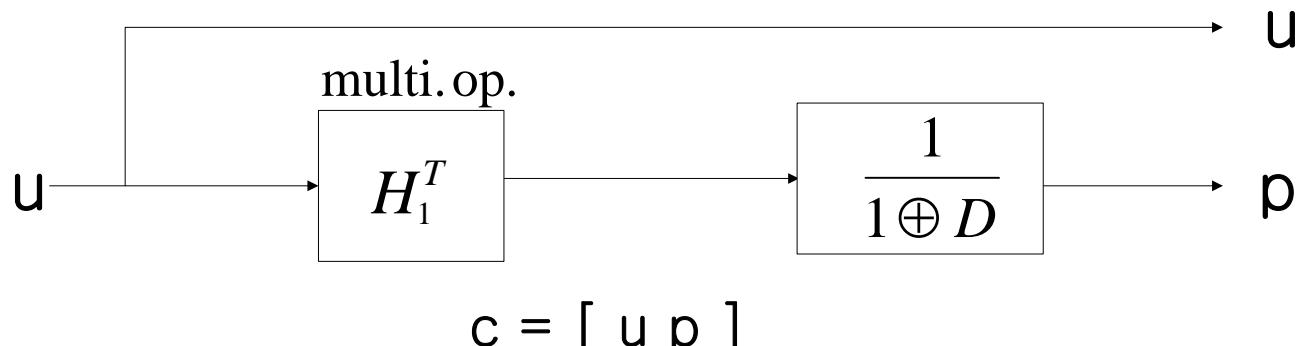
$$= \begin{bmatrix} 1 & h_{11}^T & h_{21}^T & \dots & h_{p1}^T & 1 & 1 & 1 & 1 & 1 \\ 1 & h_{12}^T & h_{22}^T & \dots & h_{p2}^T & 1 & 1 & 1 & 1 & 1 \\ \ddots & \vdots & \vdots & \ddots & \vdots & \times \mathbf{P}^T & \times & \dots & \dots & \dots \\ 1 & \vdots & \vdots & \ddots & \vdots & & & 1 & 1 & 1 \\ 1 & h_{1q}^T & h_{2q}^T & \dots & h_{pq}^T & & & & & 1 \end{bmatrix}$$

- Since $\mathbf{H}_1'^T$ is transposed form of circulant matrix, it can be constructed by shift register and interleaver.
- \mathbf{H}_2^{-T} matrix can be constructed by differential encoder.

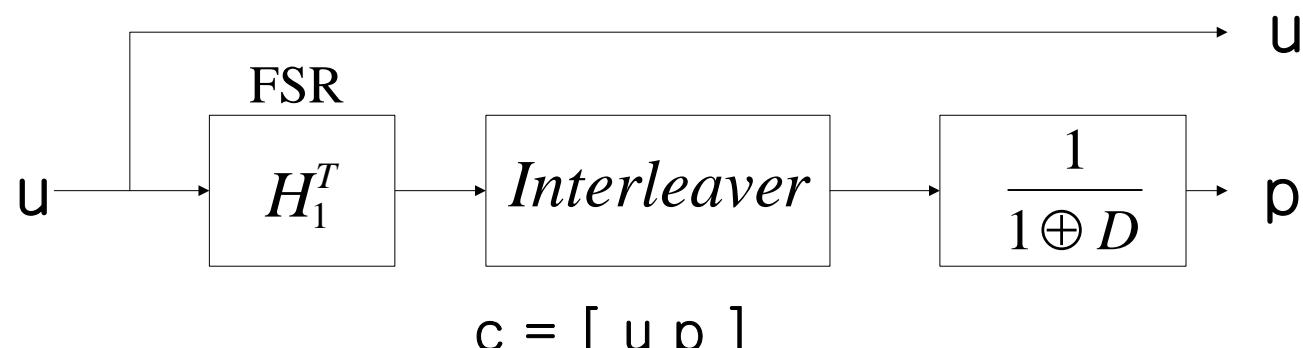
Modified Encoder Structure(4/6)

Encoder Block Diagram

a) Original Encoder



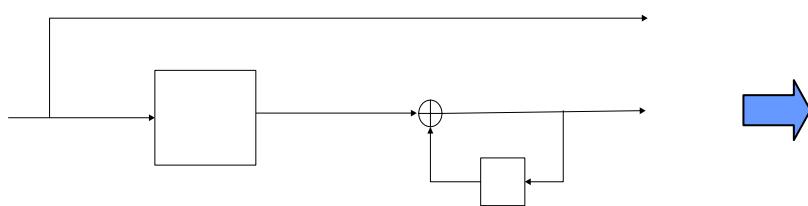
b) Modified Encoder



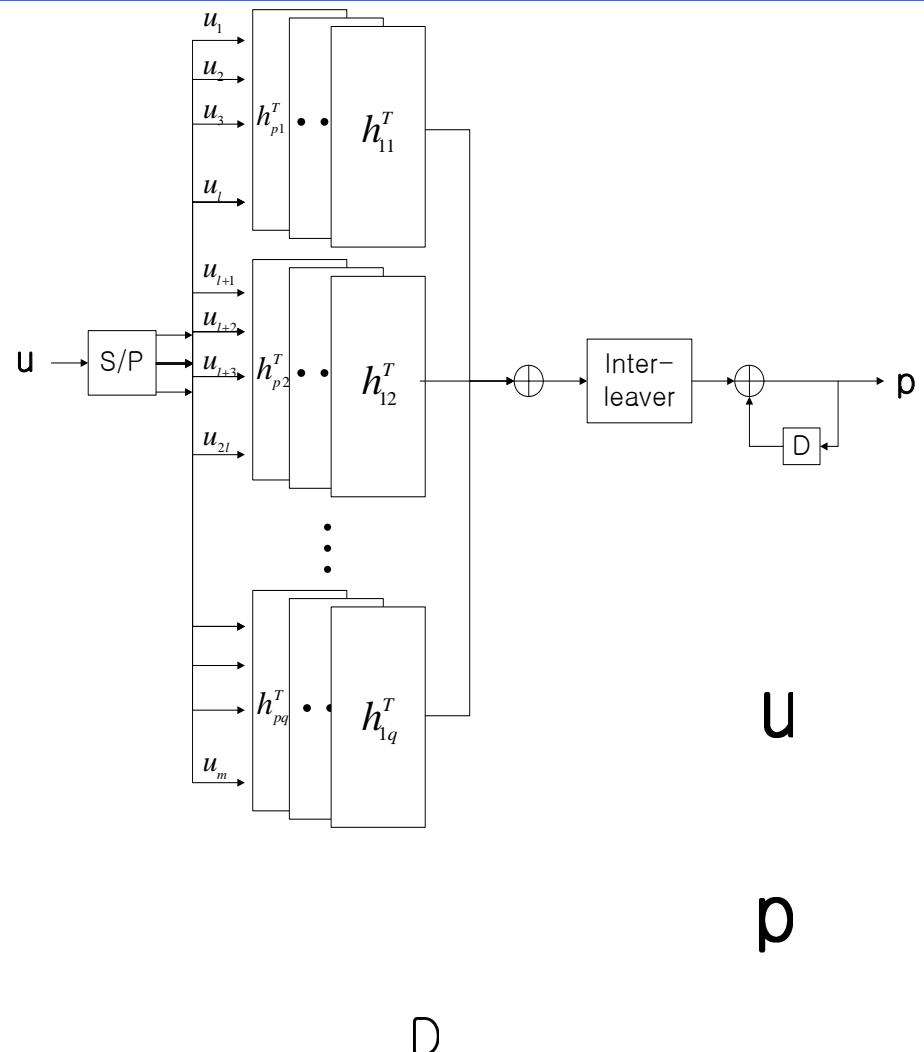
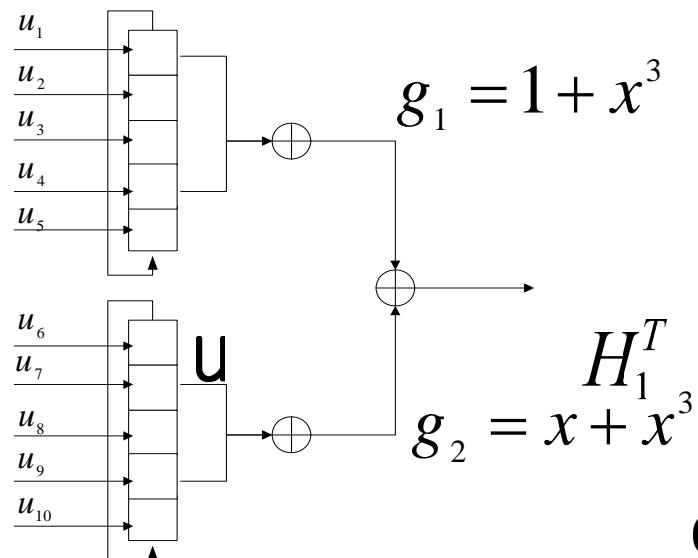
Modified Encoder Structure(5/6)

Encoder Structure

Modified encoder structure



Example of Shift Register





Modified Encoder Structure(6/6)



□ Complexity Comparison

	δ (general)	Total Computation	Total XOR	Total Memory	Computation comparision	XOR comparison ($n=1024, k=768$, $p=1$)
Matrix multiplication	≈ 0.5	$\delta k(n-k)$	$\delta k(n-k)$	$n-k$	1	•
Michael Yang	≈ 0.01	$\delta(k+1)(n-k)$	$\delta k(n-k)$	$n-k$	1/50	1
Shift Register	≈ 0.01	$\delta(k+1)(n-k)$	δkp	$n-k$	1/50	1/256
Shift Register Interleaver	≈ 0.01	$\delta(k+1)(n-k)$	δkp	$\leq 2(n-k)$	1/50	1/256

- δ : density of H_1 Matrix
- Complexity is reduced compared with matrix multiplication operation.
- In our simulation, $(n, k) = (256, 192), (512, 384), (1024, 768), (2924, 2193)$, $p=1, q=3$.



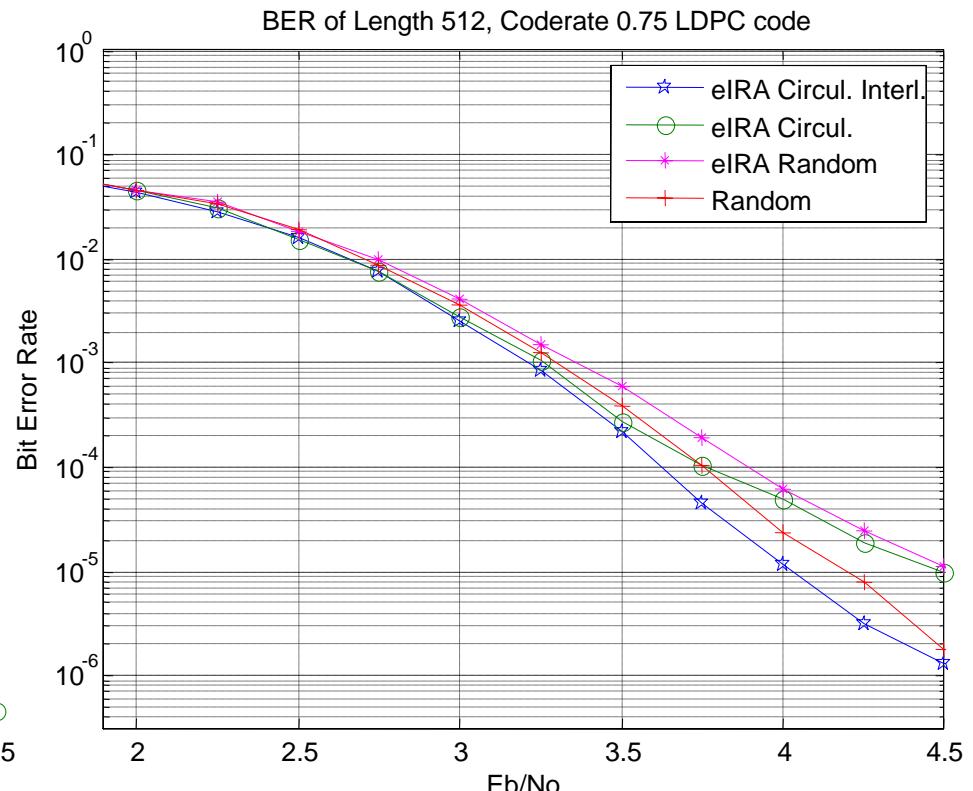
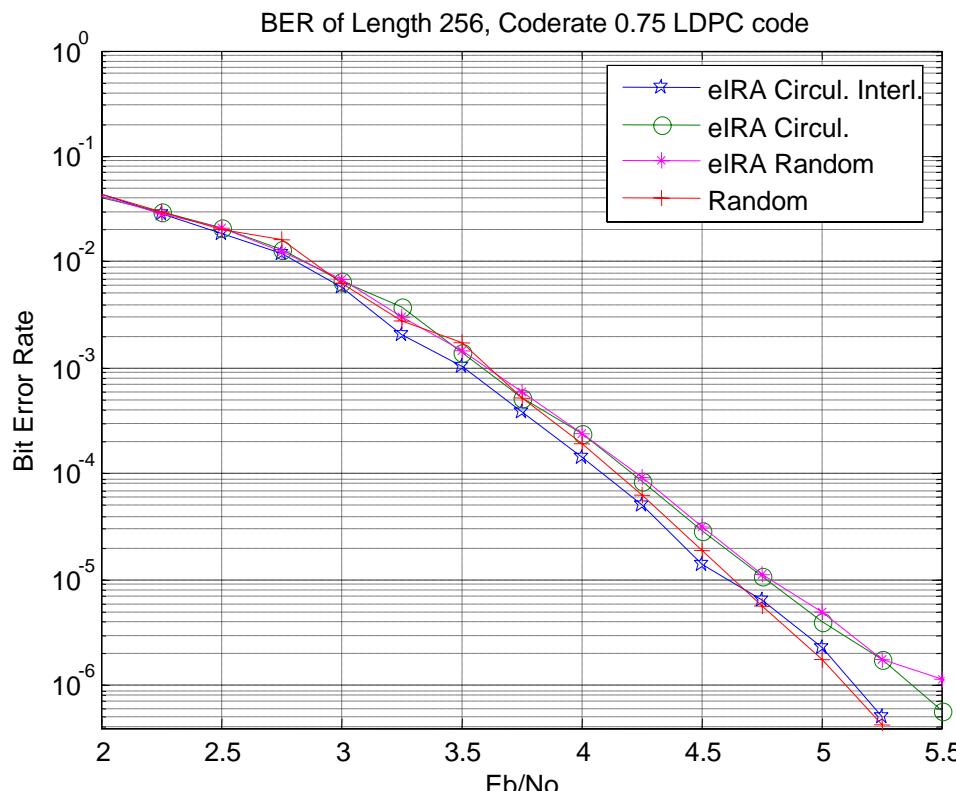
Simulation Results(1/4)

□ Simulation Environments

Decoding Algorithm	Sum-Product Algorithm
Maximum Iteration	50
Codeword Length	256, 512, 1024, 2924
Coderate	0.75
Modulation	BPSK
Channel Model	AWGN

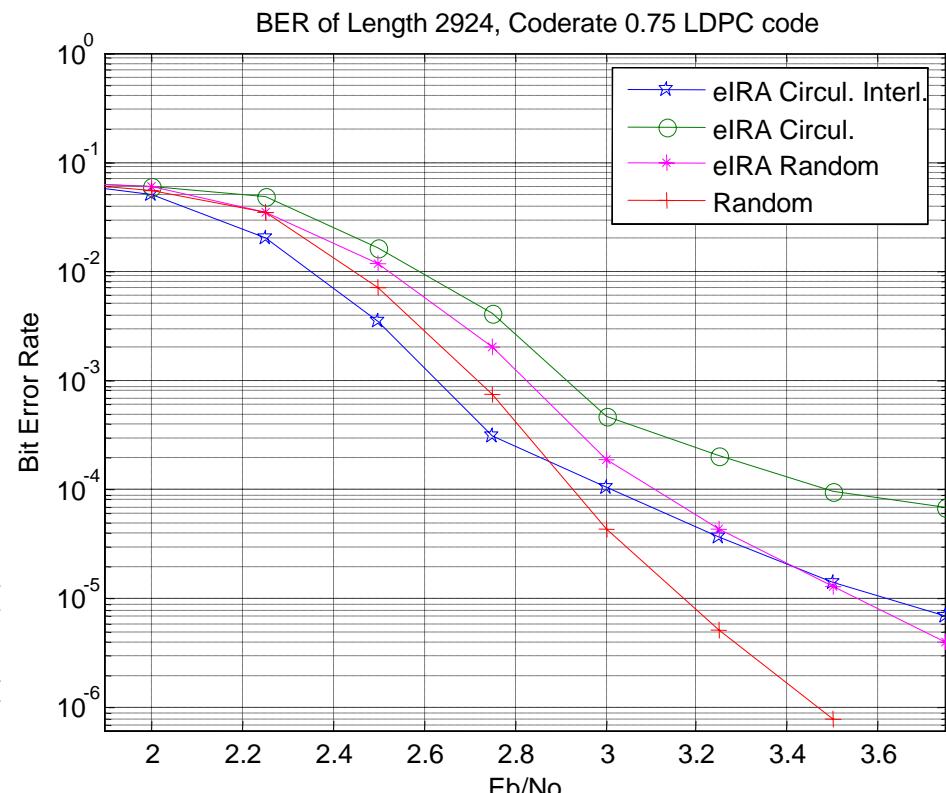
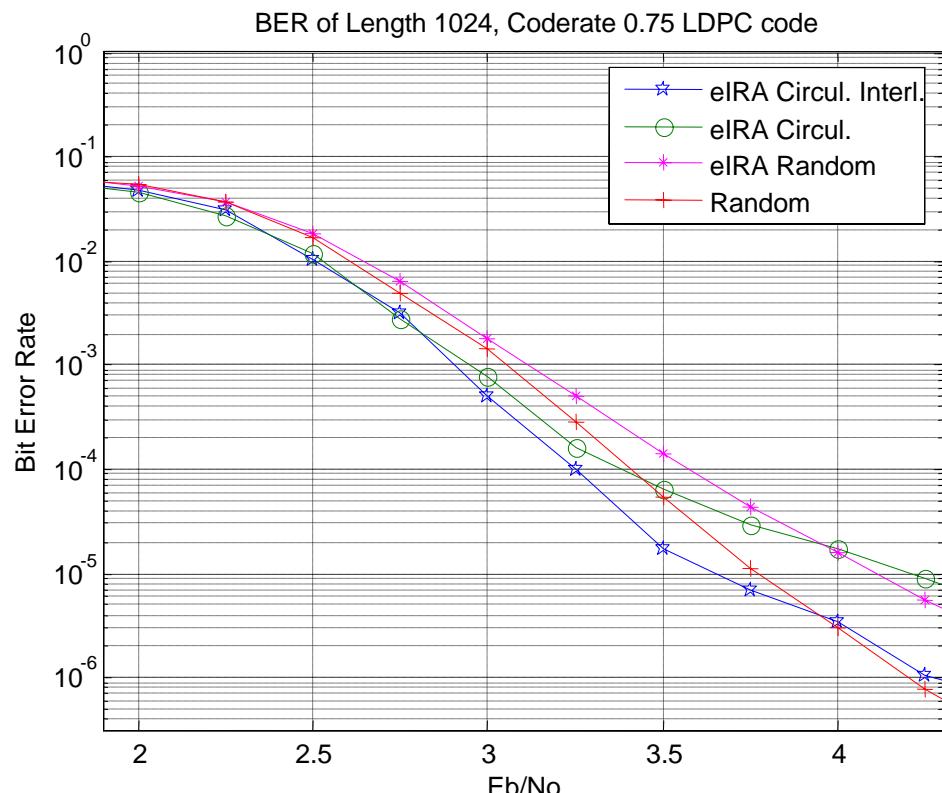
Simulation Results(2/4)

Code rate 0.75, length 256, 512



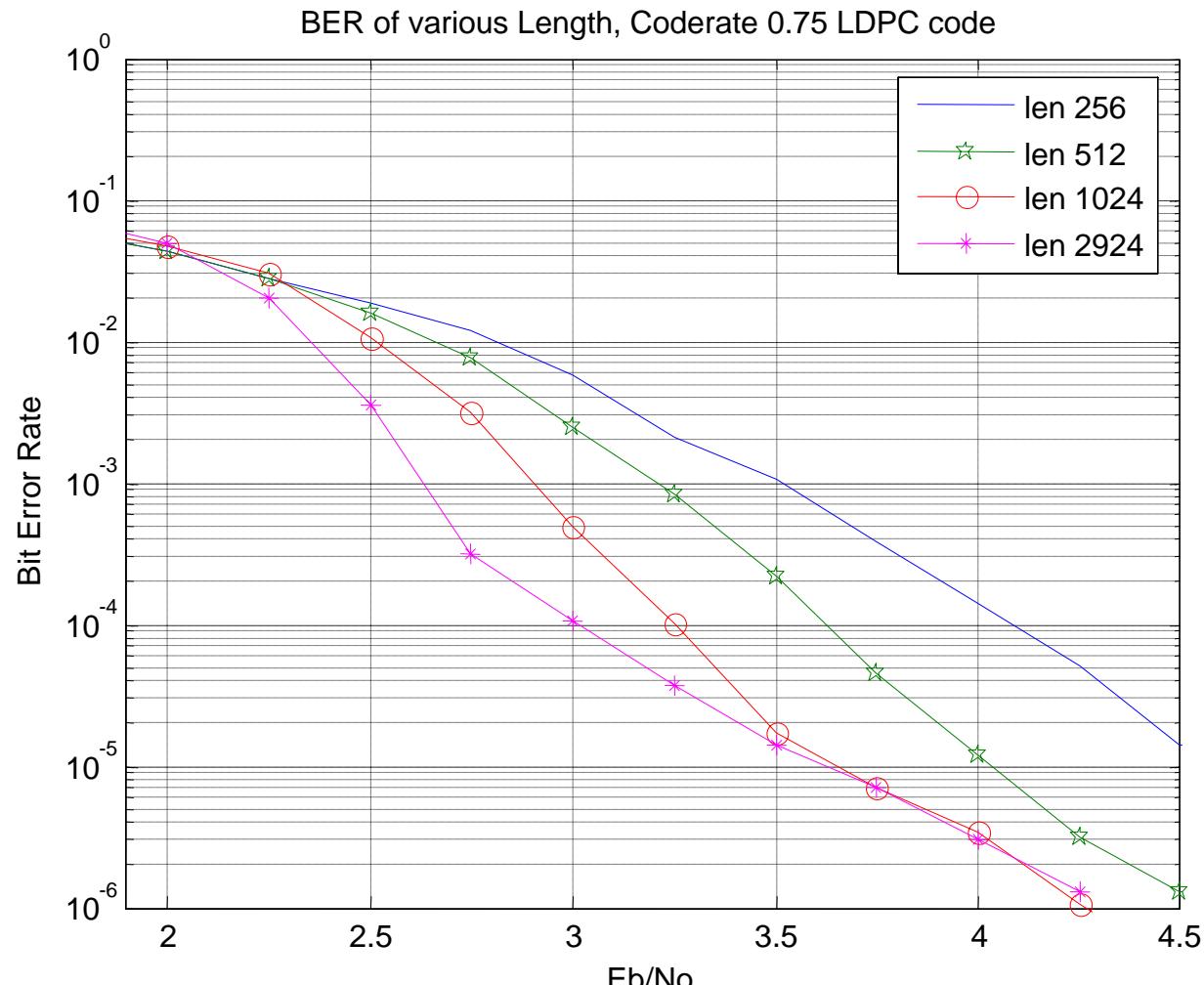
Simulation Results(3/4)

Code rate 0.75, length 1024, 2924



Simulation Results(4/4)

□ Comparison according to lengths





Concluding Remarks

□ Conclusion

- Encoder complexity is lower than Yang's method.
- Applicable to short length, high rate LDPC code

□ Future Works

- Research on generator polynomial structure to make large girth.
- Research on permutation matrix structure not to make 4-cycle after permutation.