### SISO Decoding of Coded Bi-directional Relaying System Using Block Turbo Codes

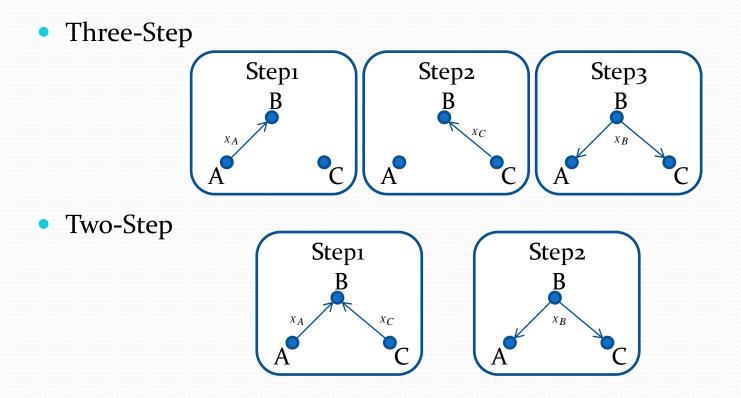
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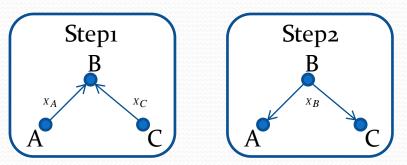


#### Coded Bi-directional Relaying



### Channel Model of Two-Step CBR

• Two-Step



• At first time slot, multiple-access channel at the relay B;

 $Y_B = X_A + X_C + Z_B$ At next time slot, it is broadcasting channel:

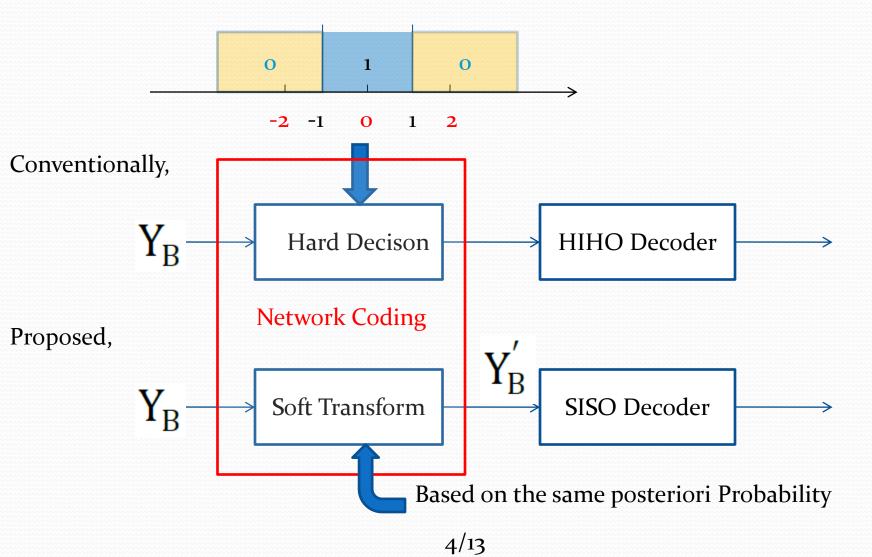
$$Y_{A} = X_{B} + Z_{A}$$

$$Y_{c} = X_{B} + Z_{c}$$

where Z is the Gaussian noise, X is modulated symbol, Y is the received value.

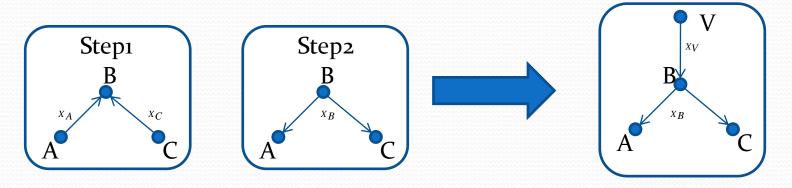
The problem of direct SISO

- If BPSK modulation is used in this system. Without Gaussian noise, Y<sub>B</sub> ∈ {-2,0,2}.
  LLR = ln(<sup>1</sup>/<sub>2</sub>\*[exp(-<sup>2</sup>/<sub>σ<sup>2</sup></sub>(1-Y<sub>B</sub>)+exp(-<sup>2</sup>/<sub>σ<sup>2</sup></sub>(1+Y<sub>B</sub>))])
  LLR ≠ <sup>2</sup>/<sub>σ<sup>2</sup></sub>Y<sub>B</sub>
- In this case, Chase-Pyndiah algorithms for Block Turbo Codes cannot be applied directly at the relay without complex derivation.



#### MAC – P2P channel

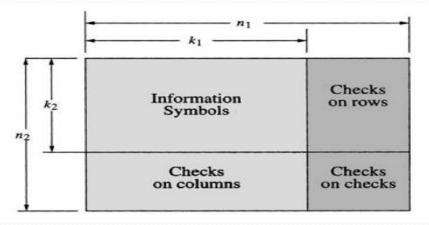
• Under the condition of the same posteriori Probability:



• The node V sent  $X_V = X_A \oplus X_C$ 

# **Block Turbo Codes**

- BTC , also known as Turbo Product Code, is one type of turbo codes, which of constituent code is block codes.
- Consider two systematic linear block codes C1 with parameters(n1,k1,d1) and C2 with parameters (n2,k2,d2). Then the parameters for the turbo product codes P=C1\*C2 is (n1\*n2,k1\*k2,d1\*d2).



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• The posteriori probability  $P(X_B = 1 | Y_B)$  can be computed as

$$P(X_{B} = 1 | Y_{B}) = \frac{\frac{1}{4} \times P(Y_{B} | X_{A} = X_{C})}{\frac{1}{4} \times [P(Y_{B} | X_{A} = X_{C}) + P(Y_{B} | X_{A} \neq X_{C})]}$$

$$P(X_{A} = 1) = P(X_{A} = -1) = \frac{1}{2}$$
$$P(X_{C} = 1) = P(X_{C} = -1) = \frac{1}{2}$$

$$P(X_{B} = 1 | Y_{B}) = \frac{\frac{1}{4} \times P(Y_{B} | X_{A} = X_{C})}{\frac{1}{4} \times [P(Y_{B} | X_{A} = X_{C}) + P(Y_{B} | X_{A} \neq X_{C})]}$$

$$P(Y_B \mid X_A = X_C = 1) = \frac{1}{\sqrt{2\pi\sigma}} e^{-\frac{1}{2\sigma^2}(Y_B - 2)^2}$$

$$P(Y_B \mid X_A = X_C = -1) = \frac{1}{\sqrt{2\pi\sigma}} e^{-\frac{1}{2\sigma^2}(Y_B + 2)^2}$$

$$P(Y_{B} \mid X_{A} = -1, X_{C} = 1) = \frac{1}{\sqrt{2\pi\sigma}} e^{-\frac{1}{2\sigma^{2}}Y_{B}^{2}}$$

$$P(Y_{B} | X_{A} = 1, X_{C} = -1) = \frac{1}{\sqrt{2\pi\sigma}} e^{-\frac{1}{2\sigma^{2}}Y_{B}^{2}}$$

• Then,

$$P(X_{B} = 1 | Y_{B}) = \frac{1}{1 + (\frac{1}{2} \times [e^{-\frac{2}{\sigma^{2}}(1 - Y_{B})} + e^{-\frac{2}{\sigma^{2}}(1 + Y_{B})}])^{-1}}$$

• The posterior probability of the soft value  $Y_{B}$  received from virtual node is

$$P(X_{B} = 1 | Y_{B}') = \frac{1}{1 + e^{(-\frac{2Y_{B}'}{\sigma^{2}})}}$$

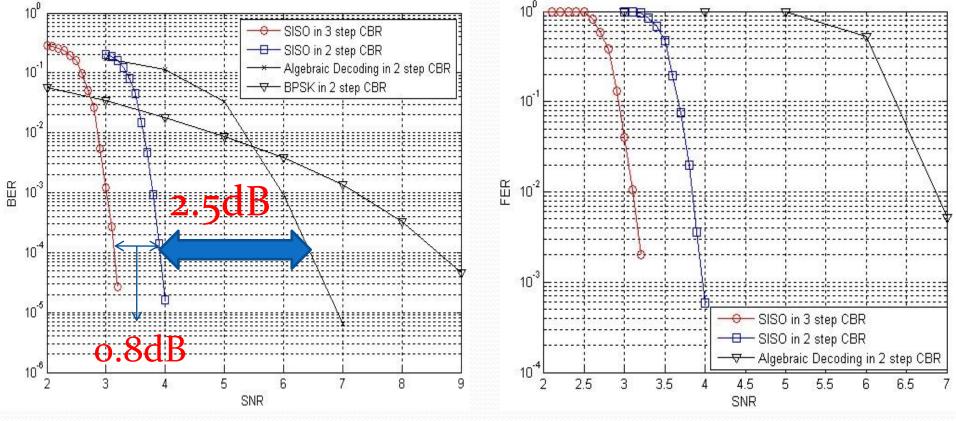
• Under the condition of the same posteriori probability  $P(X_B = 1 | Y_B) = P(X_B = 1 | Y_B)$  and using log-max approximation  $\ln(e^a + e^b) \approx \max(a, b)$ ,

$$Y_{B} = \left(\frac{\sigma^{2}}{2}\right) \times \ln\left(\frac{1}{2}\right) + \left|Y_{B}\right| - 1$$

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### Simulation Results

• Extended BCH product code (64,51,6)<sup>2</sup> is used in this simulation and Max-iteration number is set to 4.



# Conclusion

#### • Two-Step SISO scheme wins:

- ✓ 2.5dB coding gain over algebraic decoding
- Higher throughput and Lower decoding complexity at the expense of o.8dB SNR loss than Three-Step scheme

## Future Work

#### Validation of soft transformation

#### Incremental Redundancy Relaying



Q&A