

# Anti-Jamming Partially Regular LDPC Codes for Follower Jamming with Rayleigh Block Fading in Frequency Hopping Spread Spectrum

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# Outline

- 1 Introduction
- 2 System Model
- 3 AJ-PR-LDPC Codes for Follower Jamming
- 4 Simulation Result
- 5 Conclusions

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# Military Communication Scenario

- Jamming
  - **Deliberately** interferes the desired signal
  - Category
    - Full-band jamming
    - Partial band jamming: most famous
    - Follower jamming: major subject in this work
    - Smart jamming
  - Anti-jamming schemes
    - Stop-band
    - Avoiding
    - Frequency hopping spread spectrum (FHSS): considered in this work
    - Enhanced ECC (RS-concatenation, BICM-ID): major subject in this work

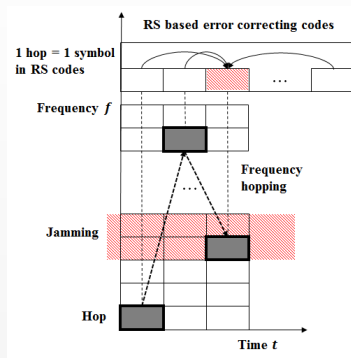


Figure: Partial band jamming and anti-jamming schemes (FHSS, ECC) in the military communication scenario

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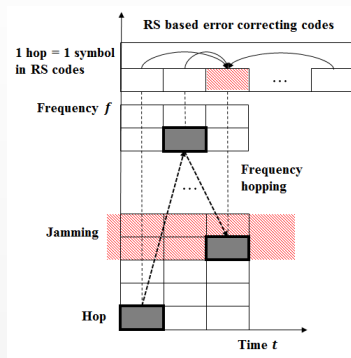


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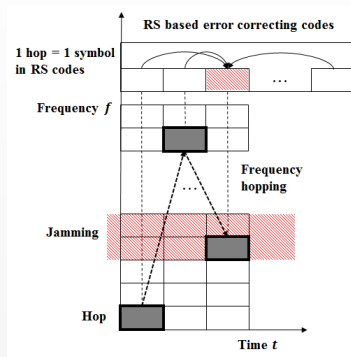


Figure: Partial band jamming and anti-jamming schemes (FHSS, ECC) in the military communication scenario

# Main Research Topic

- Follower Jamming
  - **Scan** the occupied frequency bands
  - **Send** the jamming signal in the occupied bands
  - **Use** determinant or CESM to scan the band
- Low-density Parity Check (LDPC) Codes
  - Capacity-approaching codes
  - Can be used for **special environments**
    - Block fading
    - Unequal error protection (UEP)

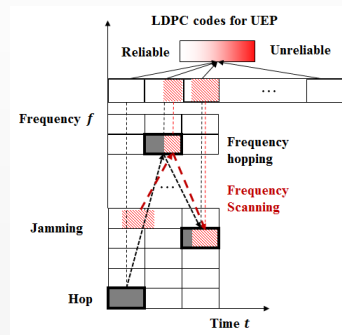


Figure: Follower jamming scenario with LDPC codes for UEP

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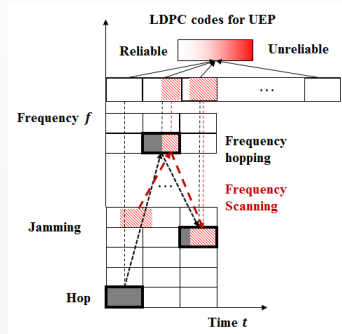


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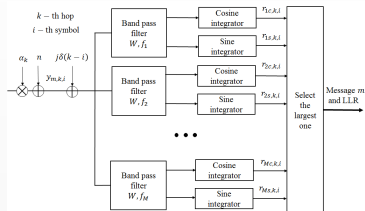


Figure: The demodulator of NC-MFSK

$$r_{mc,k,i} = \begin{cases} \alpha_k \sqrt{\mathcal{E}_{k,i}} \cos \phi + j\delta(k,i) + n, & m = \bar{m} \\ j\delta(k,i) + n, & \text{otherwise} \end{cases} \quad (1)$$

$$r_{ms,k,i} = \begin{cases} \alpha_k \sqrt{\mathcal{E}_{k,i}} \sin \phi + j\delta(k,i) + n, & m = \bar{m} \\ j\delta(k,i) + n, & \text{otherwise.} \end{cases} \quad (2)$$

$$m'_{k,i} = \operatorname{argmax}_m (\sqrt{r_{mc,k,i}^2 + r_{ms,k,i}^2}) \quad (3)$$

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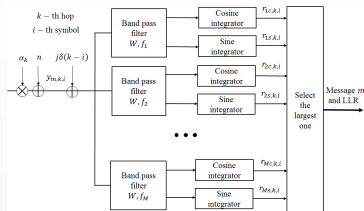


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# Follower Noise Jamming with Fixed Scan Speed

- **Geometrical Limitation** of the Follower Jamming
  - Transmission and processing time can limit the possibility of the follower jamming
  - $T_p$  (transmission) +  $T_j$  (processing)  $\leq T_h$  (hop duration)
  - $T_p = \frac{D_{tj} + D_{jr} - D_{tr}}{c}$ ,  $T_j$  can be determined by fixed speed assumption
  - $\mu = \frac{\text{(Duration of jamming exists)}}{\text{(Hop duration)}}$ , Jamming eclipse:  $[0, (1 - \mu)T_h]$

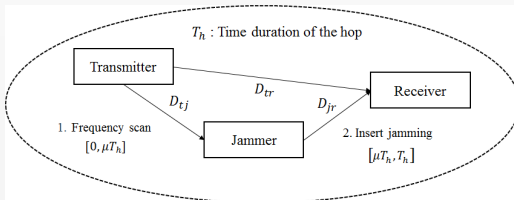


Figure: Follower jamming environment

# Follower Noise Jamming with Fixed Scan Speed (Cont')

- New Assumption: **Fixed Scan Speed  $v$** 
  - $T_j$  (processing) =  $T^*$  (inherent) +  $T_{scan}$  (scanning time)
  - $\mu \sim u[\mu_a, \min(\mu_b, 1)]$ ,  $\mu_a$ (initial) =  $\frac{T_p + T^*}{T_h}$ ,  $\mu_b$ (end) =  $\mu_a + \frac{N_{fr}}{vT_h}$ .
  - $\rho$  (prob. that jamming exists in a hop) =  $\min\left\{\frac{1 - \min(1, \mu_a)}{\mu_b - \mu_a}, 1\right\}$

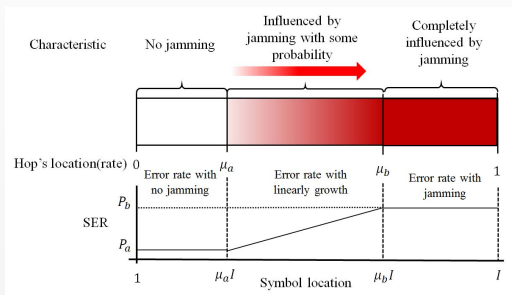


Figure: Average symbol error rate of the hop in the presence of the follower jamming

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# $(\lambda, d_c, \mathbf{d}_v)$ AJ-PR-LDPC codes

- Parameters  $(\lambda, d_c, \mathbf{d}_v)$ 
  - Location vector**  $\lambda = [\lambda_1, \dots, \lambda_{|\lambda|}]$ ,  
 $\sum \lambda_i = 1$
  - Variable node degree**  
 $\mathbf{d}_v = [d_{v,1}, \dots, d_{v,|\lambda|}]$  with  $r(\lambda \cdot \mathbf{d}_v) = d_c$

- AJ-PR-LDPC codes
  - Constant check node degree  $d_c$  on the each row
  - Each  $v_j$ ,  $j$ -th column of each hop whose size is  $B$ , has to satisfy

$$wt(v_j) = d_{v,k}, j \in \left[ B \sum_{i=1}^{k-1} \lambda_i, B \sum_{i=1}^k \lambda_i \right] \quad (4)$$

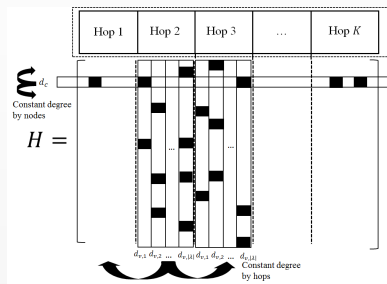


Figure: Parity check matrix of AJ-PR-LDPC codes

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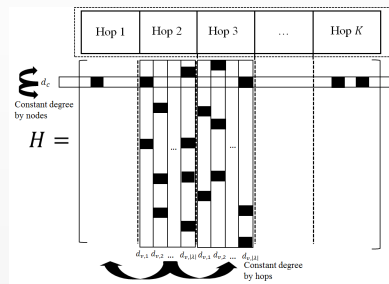


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# Simplified Channel Model for Density Evolution

- Density Evolution and Simplified Channel Model
  - Mathematical tools for analyzing **asymptotic** performance of the LDPC codes
  - **Difficult** to apply the error channel with linearly growth and general LDPC codes.
  - **Alternative solution**: simplified channel model with stair-case erasure and partially regular structure.

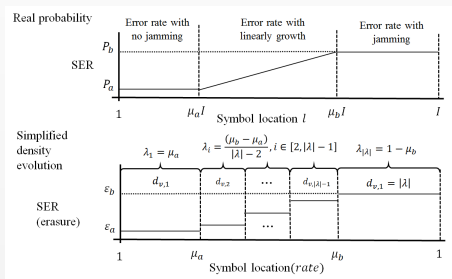


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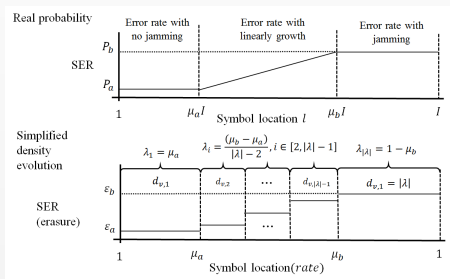


Figure: Simplified channel model

# Corresponding Density Evolution and Construction Algorithm

- Corresponding Density Evolution of Erasure Channel

$$\epsilon_i = (\epsilon_b - \epsilon_a) \frac{i-1}{|\lambda|-1} + \epsilon_a, i \in [1, |\lambda|] \quad (5)$$

$$q_{l+1} = 1 - \left( 1 - \sum_{i=1}^{|\lambda|} \lambda_i p_{l,i} \right)^{d_c-1} \quad (6)$$

$$p_{l+1,i} = \epsilon_i q_{l+1}^{d_v,i-1} \quad (7)$$

- Construction Algorithm of the Parity Check Matrix

- Search for the all degree values less than  $d_{v,max}$  and  $d_{c,max}$
- Select the remaining degree values converged to 0 with increasing erasure probability
- Construct  $H$  by partially regular PEG (slight modification of regular PEG)

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# Simulation Criteria

- System Parameters

- SFH and NC-MFSK with Rayleigh block fading
- $M = 2, 4, 8,$  and  $16$
- Hop size:  $192[\text{bits}]$

- Jamming Parameters

- Slow scan:  $\mu_a = \frac{3}{8}, \mu_b = \frac{11}{8}, \rho = \frac{5}{8}, \frac{E_b}{N_j} = -50[\text{dB}]$
- Fast scan:  $\mu_a = \frac{3}{8}, \mu_b = \frac{7}{8}, \rho = 1, \frac{E_b}{N_j} = -50[\text{dB}]$

- Codes in the Simulation

- $(2304, 1152)$  LDPC codes in IEEE 802.16e standard
- $(2304, 1152)$  AJ-PR-LDPC (proposed) codes with  $\lambda = (\frac{5}{8}, \frac{2}{8}, \frac{1}{8}),$   
 $d_v = (2, 3, 4),$  and  $d_c = 5$



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# Simulation results

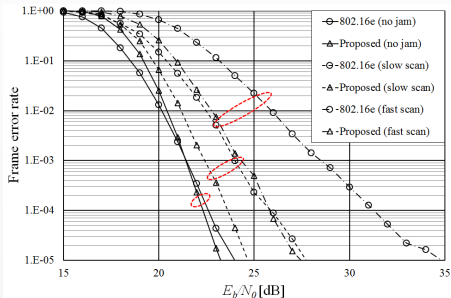


Figure: Decoding performance when  $M = 2$

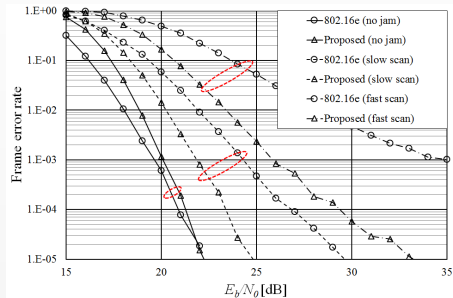


Figure: Decoding performance when  $M = 4$

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