

A generalized Milewski construction for perfect sequences for APSK+ constellations

송민규, 송홍엽
(연세대학교)

2018년 부호 및 정보이론 연구회 3차 워크숍



Autocorrelation & perfect sequence



- Let $\mathbf{x} = \{x(n)\}_{n=0}^{L-1}$ be a complex-valued sequence of length L .

The (periodic) autocorrelation of \mathbf{x} at time shift τ is

$$C_{\mathbf{x}}(\tau) = \sum_{n=0}^{L-1} x(n + \tau)x^*(n).$$

where $n + \tau$ is computed over the integers modulo L .

- A sequence is called **perfect** if

$$C_{\mathbf{x}}(\tau) = E\delta(\tau) = \begin{cases} E, & \tau \equiv 0 \pmod{L} \\ 0, & \tau \not\equiv 0 \pmod{L}. \end{cases}$$

where $E = C_{\mathbf{x}}(0)$

DFT of a perfect sequence

$$\begin{aligned}
 \Im\{\mathcal{C}_x(\tau)\}(k) &= \sum_{\tau=0}^{L-1} \left[\sum_{n=0}^{L-1} x(n + \tau) x^*(n) \right] e^{-j2\pi k \tau} \\
 &= \sum_{n=0}^{L-1} \left[\sum_{\tau=0}^{L-1} x(n + \tau) e^{-j2\pi k(n+\tau)} \right] x^*(n) e^{j2\pi kn} \\
 &= X(k) \left[\sum_{n=0}^{L-1} x(n) e^{-j2\pi kn} \right]^* = X(k)X^*(k) = |X(k)|^2
 \end{aligned}$$

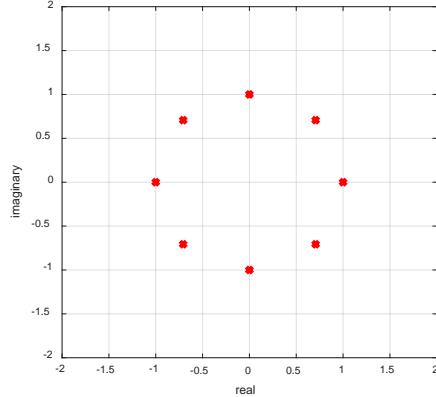
$\mathcal{C}_x(\tau) = E\delta(\tau)$ if and only if $|X(k)|^2$ is a constant for any k .

x is a **perfect** sequence of period L **if and only if**

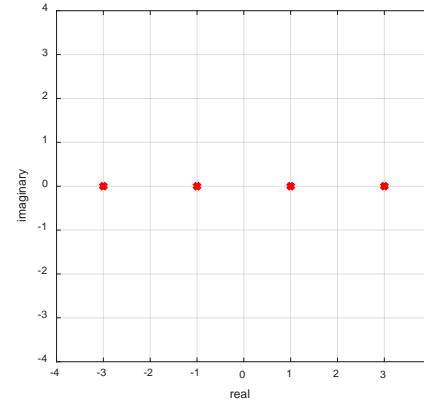
its **DFT** has **flat magnitude**.

Constellations (alphabets)

Phase shift keying (PSK) → **polyphase** Amplitude shift keying (ASK)

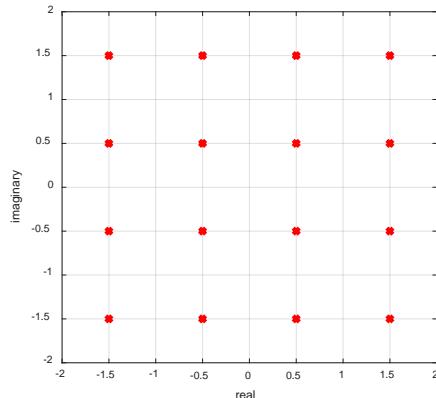


Phase only
well studied



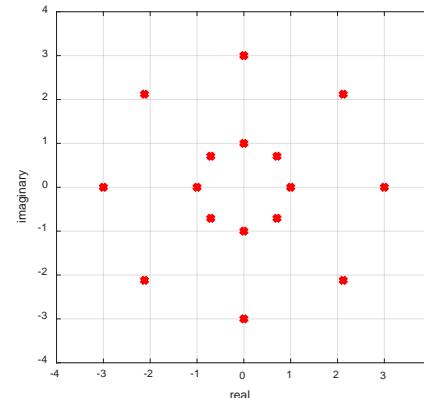
real amplitude only
few results

Quadrature-amplitude modulation (QAM)



Normally square
studied

Amplitude & phase shift keying (APSK)

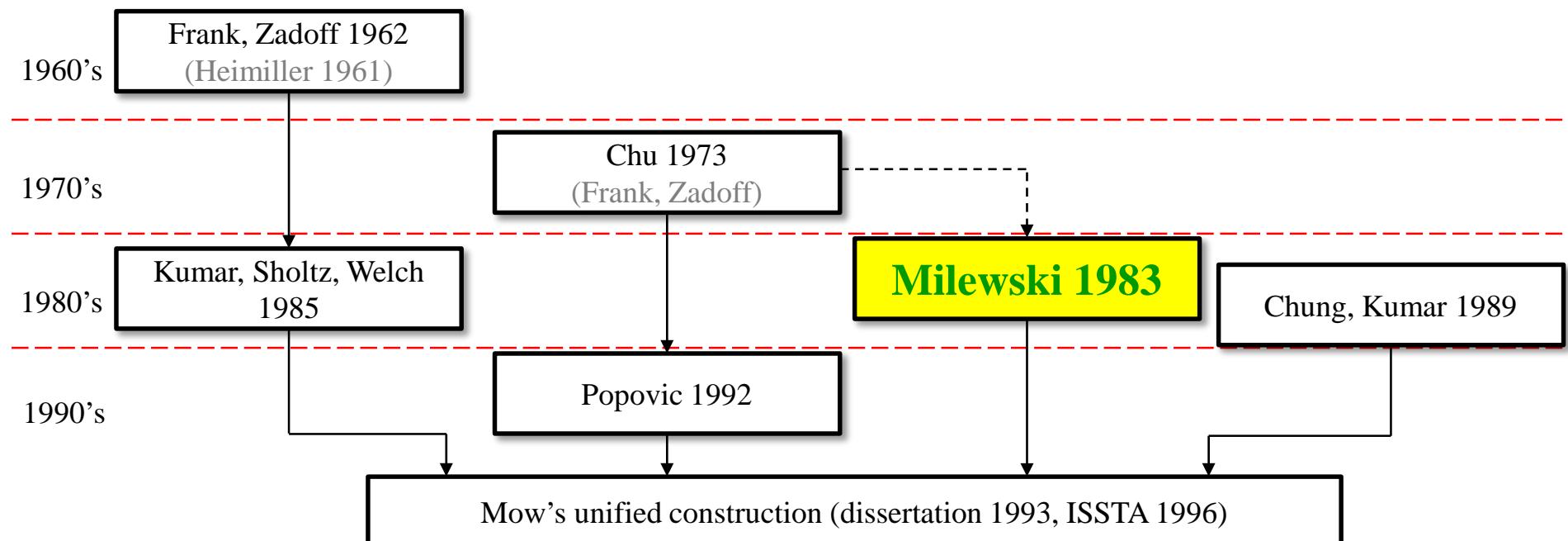


Both Phase and Amplitude
few results

Adding the zero point ⇒ PSK+, ASK+, QAM+, and APSK+

The Milewski construction

- Originally proposed to construct perfect polyphase sequences from those of shorter period
 - Well-known case: using the Zadoff-Chu sequence as the shorter one



- Brief history of constructing perfect polyphase sequences -



Interleaved sequence

Let $S = \{s_0, s_1\}$ be a set of 2 sequences of length 3, where

$$s_0 = \{a, b, c\}, s_1 = \{d, e, f\}.$$

Write an array as

$$[s_0, s_1] = \begin{bmatrix} a & d \\ b & e \\ c & f \end{bmatrix},$$

and then, by reading the array row-by-row,
we have

$$s = \{a, d, b, e, c, f\}.$$

The sequence s is called an interleaved sequence.



Cyclic shift of interleaved sequence



$$s = \{a, d, b, e, c, f\}$$



1 left-cyclic shift

$$\{d, b, e, c, f, a\}$$



1 left-cyclic shift

$$\{b, e, c, f, a, d\}$$

$$\begin{bmatrix} a & d \\ b & e \\ c & f \end{bmatrix}$$



$$\begin{bmatrix} d & b \\ e & c \\ f & a \end{bmatrix}$$



$$\begin{bmatrix} b & e \\ c & f \\ a & d \end{bmatrix}$$



The original Milewski construction

period: $m^1 \rightarrow m^{2K+1}$



perfect polyphase
sequence

A positive
integer

$$\alpha = \{\alpha(n)\}_{n=0}^{m-1}$$

K



Output
perfect polyphase sequence

$$s = \{s(n)\}_{n=0}^{m^{2K+1}-1}$$

where

$$s(n) = \alpha(q)\omega^{qr}$$

where $n = q\mathbf{m^K} + r$.

and

$$\omega = e^{-\frac{j2\pi}{m^{1+K}}}$$

The original Milewski construction

period: $m^1 \rightarrow m^{1+2K}$

perfect polyphase sequence

$$\alpha = \{\alpha(n)\}_{n=0}^{m-1}$$

A positive integer
 K



Output perfect polyphase sequence
 $s = \{s(n)\}_{n=0}^{m^{2K+1}-1}$

where

$$s(n) = \alpha(q)\omega^{qr}$$

where $n = q\mathbf{m^K} + r$.

and

$$\omega = e^{-\frac{j2\pi}{m^{K+1}}}.$$

$m \cdot m^K \times m^K$ array form of s

Input sequence of period m

$\alpha(0)$	$\times 1$	$\alpha(0)$	$\times 1$	\dots	$\alpha(0)$	$\times 1$
$\alpha(1)$	$\times 1$	$\alpha(1)$	$\times \omega$	\dots	$\alpha(1)$	$\times \omega^{m^K-1}$
$\alpha(2)$	$\times 1$	$\alpha(2)$	$\times \omega^2$	\dots	$\alpha(2)$	$\times (\omega^{m^K-1})^2$
\vdots		\vdots		\ddots	\vdots	\vdots
$\alpha(m-1)$	$\times 1$	$\alpha(m-1)$	$\times \omega^{m^K-1}$	\dots	$\alpha(m-1)$	$\times (\omega^{m^K-1})^{m^K-1}$
\vdots		\vdots		\ddots	\vdots	\vdots
$\alpha(0)$	$\times 1$	$\alpha(0)$	$\times \omega^{m^K}$	\dots	$\alpha(0)$	$\times (\omega^{m^K-1})^{m^K}$
$\alpha(1)$	$\times 1$	$\alpha(1)$	$\times \omega^{m^K+1}$	\dots	$\alpha(1)$	$\times (\omega^{m^K-1})^{m^K+1}$
$\alpha(2)$	$\times 1$	$\alpha(2)$	$\times \omega^{m^K+2}$	\dots	$\alpha(2)$	$\times (\omega^{m^K-1})^{m^K+2}$
\vdots		\vdots		\ddots	\vdots	\vdots
$\alpha(m-1)$	$\times 1$	$\alpha(m-1)$	$\times \omega^{m^K+1-1}$	\dots	$\alpha(m-1)$	$\times (\omega^{m^K-1})^{m^K+1-1}$



Our generalization

perfect polyphase sequence

$$\alpha = \{\alpha(n)\}_{n=0}^{m-1}$$

A positive integer

$$K$$

Milewski Construction

Output

perfect polyphase sequence
 $s = \{s(n)\}_{n=0}^{m^{2K+1}-1}$

where

$$s(n) = \alpha(q)\omega^{qr}$$

with $n = q\mathbf{m}^K + r$, and

$$\omega = e^{-\frac{j2\pi}{m^{K+1}}} = e^{-\frac{j2\pi}{m(m^K)}}$$

Perfect sequence
not necessarily polyphase

$$\alpha = \{\alpha(n)\}_{n=0}^{m-1}$$

A positive integer

$$N$$

Generalized Milewski Construction

Output

perfect sequence
 $s = \{s(n)\}_{n=0}^{mN^2-1}$

where

$$s(n) = \mu(r)\alpha(q)\omega^{qr}$$

with $n = q\mathbf{N} + r$, and

$$\omega = e^{-\frac{j2\pi}{mN}}$$



Array Form

Assume that μ is the all-1 sequence,

Column index $r = 0, 1, 2, \dots, N - 1$

Row index $q = 0, 1, 2, \dots, mN - 1$

$\alpha(0)$	$\times (\omega^0)^0$	$\alpha(0)$	$\times (\omega^1)^0$	\dots	$\alpha(0)$	$\times (\omega^{N-1})^0$
$\alpha(1)$	$\times (\omega^0)^1$	$\alpha(1)$	$\times (\omega^1)^1$	\dots	$\alpha(1)$	$\times (\omega^{N-1})^1$
$\alpha(2)$	$\times (\omega^0)^2$	$\alpha(2)$	$\times (\omega^1)^2$	\dots	$\alpha(2)$	$\times (\omega^{N-1})^2$
\vdots		\vdots		\ddots		\vdots
$\alpha(m-1) \times (\omega^0)^{m-1}$		$\alpha(m-1) \times (\omega^1)^{m-1}$		\dots	$\alpha(m-1) \times (\omega^{N-1})^{m-1}$	
\vdots		\vdots		\ddots		\vdots
$\alpha(0)$	$\times (\omega^0)^{m(N-1)}$	$\alpha(0)$	$\times (\omega^1)^{m(N-1)}$	\dots	$\alpha(0)$	$\times (\omega^{N-1})^{m(N-1)}$
$\alpha(1)$	$\times (\omega^0)^{m(N-1)+1}$	$\alpha(1)$	$\times (\omega^1)^{m(N-1)+1}$	\dots	$\alpha(1)$	$\times (\omega^{N-1})^{m(N-1)+1}$
$\alpha(2)$	$\times (\omega^0)^{m(N-1)+2}$	$\alpha(2)$	$\times (\omega^1)^{m(N-1)+2}$	\dots	$\alpha(2)$	$\times (\omega^{N-1})^{m(N-1)+2}$
\vdots		\vdots		\ddots		\vdots
$\alpha(m-1) \times (\omega^0)^{mN-1}$		$\alpha(m-1) \times (\omega^1)^{mN-1}$		\dots	$\alpha(m-1) \times (\omega^{N-1})^{mN-1}$	

$$\hat{\otimes} \omega = e^{-\frac{j2\pi}{mN}}$$

Proof when μ is all-1 sequence

$$\begin{array}{c}
 \left. \begin{array}{ccccccc}
 \alpha(0) & \times (\omega^0)^0 & \alpha(0) & \times (\omega^1)^0 & \cdots & \alpha(0) & \times (\omega^{N-1})^0 \\
 \alpha(1) & \times (\omega^0)^1 & \alpha(1) & \times (\omega^1)^1 & \cdots & \alpha(1) & \times (\omega^{N-1})^1 \\
 \alpha(2) & \times (\omega^0)^2 & \alpha(2) & \times (\omega^1)^2 & \cdots & \alpha(2) & \times (\omega^{N-1})^2 \\
 \vdots & & \vdots & & \ddots & & \vdots \\
 \alpha(m-1) \times (\omega^0)^{m-1} & \alpha(m-1) \times (\omega^1)^{m-1} & \cdots & \alpha(m-1) \times (\omega^{N-1})^{m-1} \\
 \vdots & \vdots & \ddots & & \vdots & & \vdots \\
 \alpha(0) & \times (\omega^0)^{m(N-1)} & \alpha(0) & \times (\omega^1)^{m(N-1)} & \cdots & \alpha(0) & \times (\omega^{N-1})^{m(N-1)} \\
 \alpha(1) & \times (\omega^0)^{m(N-1)+1} & \alpha(1) & \times (\omega^1)^{m(N-1)+1} & \cdots & \alpha(1) & \times (\omega^{N-1})^{m(N-1)+1} \\
 \alpha(2) & \times (\omega^0)^{m(N-1)+2} & \alpha(2) & \times (\omega^1)^{m(N-1)+2} & \cdots & \alpha(2) & \times (\omega^{N-1})^{m(N-1)+2} \\
 \vdots & \vdots & \ddots & & \vdots & & \vdots \\
 \alpha(m-1) \times (\omega^0)^{mN-1} & \alpha(m-1) \times (\omega^1)^{mN-1} & \cdots & \alpha(m-1) \times (\omega^{N-1})^{mN-1}
 \end{array} \right\}$$

$$\triangleq (s_0, s_1, s_2, \dots, s_{N-1})$$

where

$$s_r = \{s_r(q) = \alpha(q)(\omega^r)^q\}_{q=0}^{mN-1}$$

is the r -th column

Note that the DFT is linear.



Proof - continued

$$(s_0, s_1, s_2, \dots, s_{N-1}) = \underbrace{(s_0, 0, 0, \dots, 0)}_{\downarrow} + \dots + (0, 0, \dots, 0, s_{N-1})$$

$\alpha(0)$	$\times (\omega^0)^0$	0	\cdots	0
$\alpha(1)$	$\times (\omega^0)^1$	0	\cdots	0
$\alpha(2)$	$\times (\omega^0)^2$	0	\cdots	0
\vdots		\vdots	\ddots	\vdots
$\alpha(m-1) \times (\omega^0)^{m-1}$		0	\cdots	0
\vdots		\vdots	\ddots	\vdots
$\alpha(0)$	$\times (\omega^0)^{m(N-1)}$	0	\cdots	0
$\alpha(1)$	$\times (\omega^0)^{m(N-1)+1}$	0	\cdots	0
$\alpha(2)$	$\times (\omega^0)^{m(N-1)+2}$	0	\cdots	0
\vdots		\vdots	\ddots	\vdots
$\alpha(m-1) \times (\omega^0)^{mN-1}$		0	\cdots	0



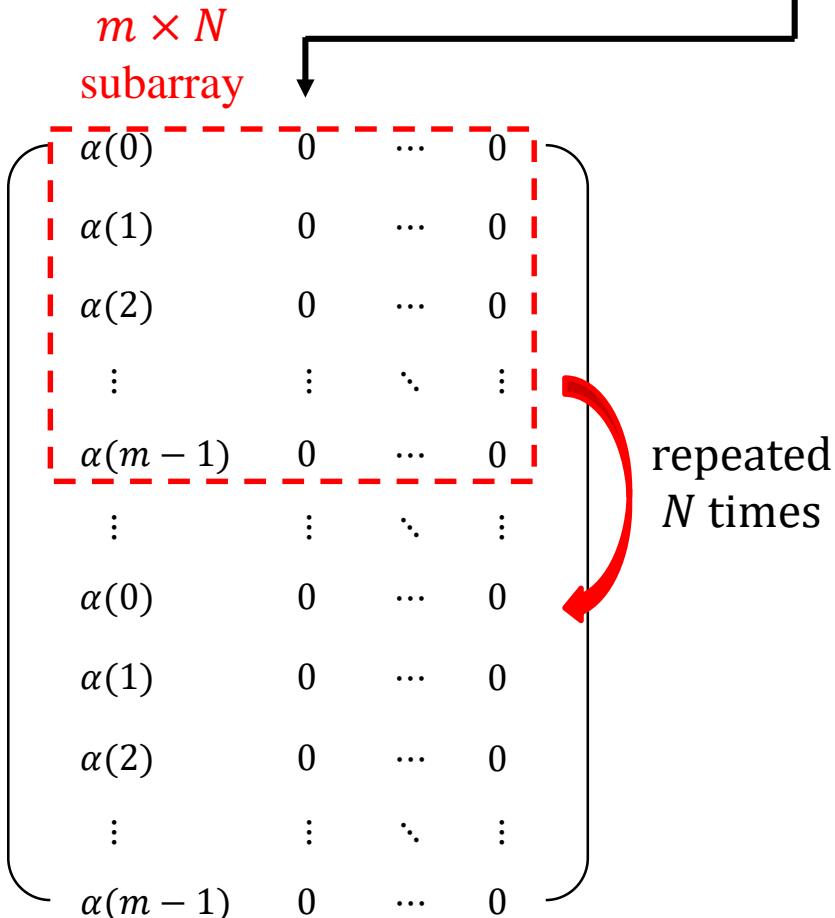
Proof - continued

$$(s_0, s_1, s_2, \dots, s_{N-1}) = \underbrace{(s_0, 0, 0, \dots, 0)}_{\downarrow} + \dots + (0, 0, \dots, 0, s_{N-1})$$

$\alpha(0)$	0	\cdots	0
$\alpha(1)$	0	\cdots	0
$\alpha(2)$	0	\cdots	0
\vdots	\vdots	\ddots	\vdots
$\alpha(m-1)$	0	\cdots	0
\vdots	\vdots	\ddots	\vdots
$\alpha(0)$	0	\cdots	0
$\alpha(1)$	0	\cdots	0
$\alpha(2)$	0	\cdots	0
\vdots	\vdots	\ddots	\vdots
$\alpha(m-1)$	0	\cdots	0

Proof - continued

$$(s_0, s_1, s_2, \dots, s_{N-1}) = \underline{(s_0, 0, 0, \dots, 0)} + \dots + (\underline{0, 0, \dots, 0}, s_{N-1})$$



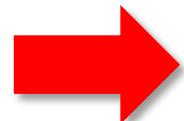


Proof - continued



$m \times N$ array

$$\begin{bmatrix} \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \end{bmatrix}$$



Interleaved sequence

$$\alpha(0) 0 \cdots 0 \quad \underbrace{\alpha(1) 0 \cdots 0}_{N-1 \text{ times}} \quad \underbrace{\alpha(2) 0 \cdots 0}_{N-1 \text{ times}} \quad \cdots \quad \underbrace{\alpha(m-1) 0 \cdots 0}_{N-1 \text{ times}}$$

Case 1) any non-zero term meets another non-zero term

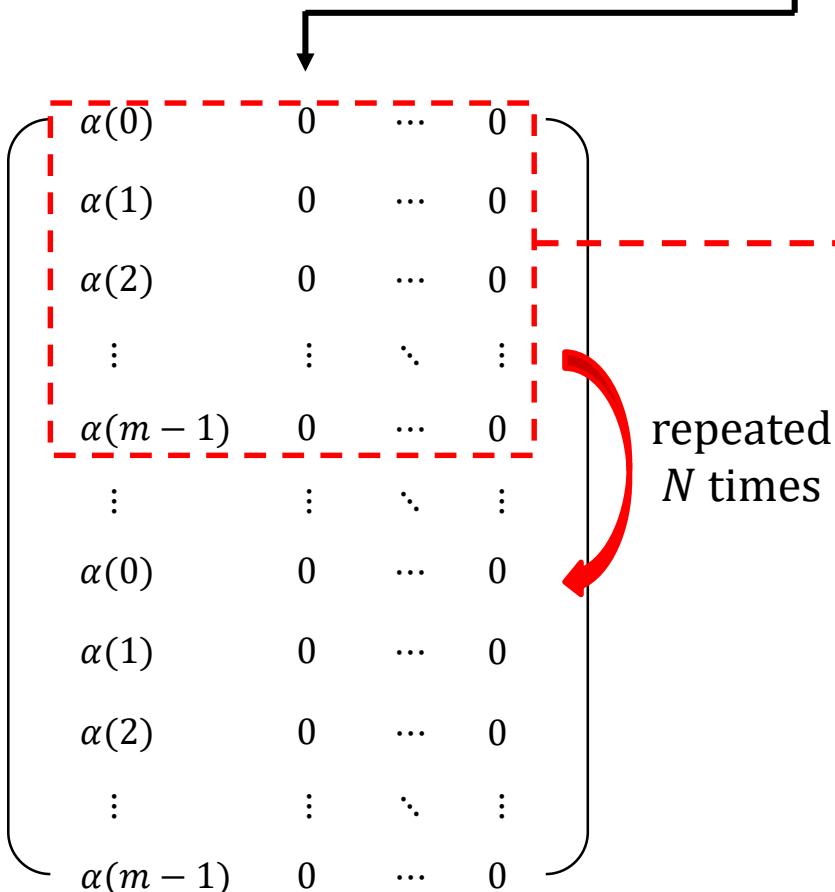
$$\begin{array}{ccccccccccccc} \alpha(\tau) & 0 & \cdots & 0 & \alpha(\tau+1) & 0 & \cdots & 0 & \alpha(\tau+2) & 0 & \cdots & 0 & \cdots & \alpha(\tau-1) & 0 & \cdots & 0 \\ \alpha^*(0) & 0 & \cdots & 0 & \alpha^*(1) & 0 & \cdots & 0 & \alpha^*(2) & 0 & \cdots & 0 & \cdots & \alpha^*(m-1) & 0 & \cdots & 0 \end{array}$$

Case 2) any non-zero term meets zero term.

$$\begin{array}{ccccccccccccc} \alpha(\tau) & 0 & \cdots & 0 & \alpha(\tau+1) & 0 & \cdots & 0 & \alpha(\tau+2) & 0 & \cdots & 0 & \cdots & \alpha(\tau-1) & 0 & \cdots & 0 \\ 0 & \alpha^*(0) & 0 & \cdots & 0 & \alpha^*(1) & 0 & \cdots & 0 & \alpha^*(2) & 0 & \cdots & 0 & \cdots & 0 & \alpha^*(m-1) & 0 & \cdots \end{array}$$

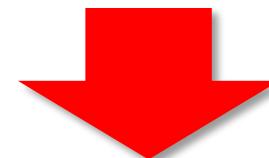
Proof - continued

$$(s_0, s_1, s_2, \dots, s_{N-1}) = \underbrace{(s_0, 0, 0, \dots, 0)} + \dots + (0, 0, \dots, 0, s_{N-1})$$

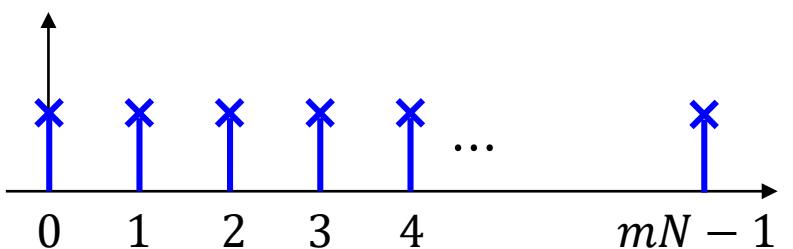


A perfect sequence of period mN

repeated
 N times

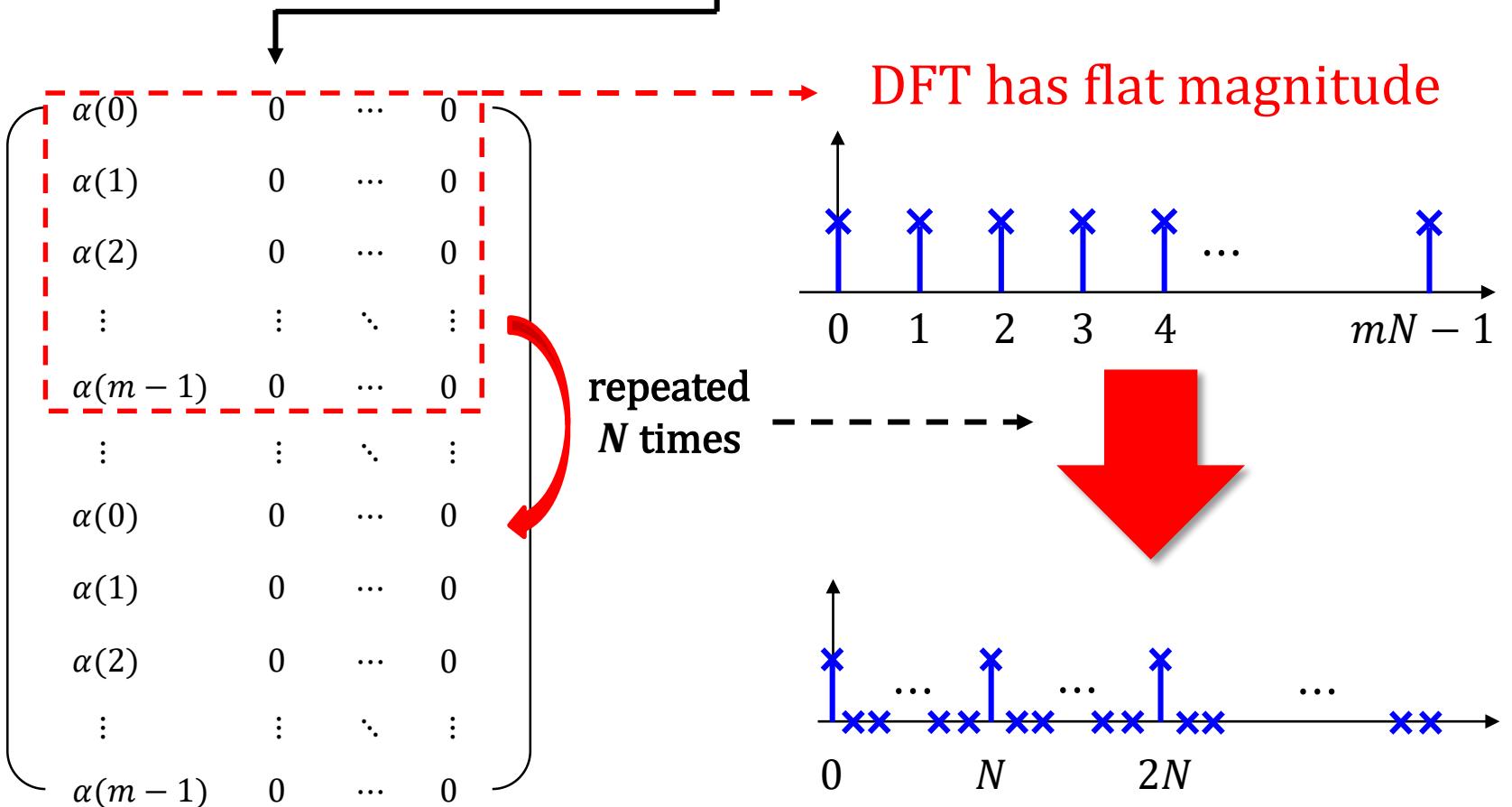


Its DFT has flat magnitude



Proof - continued

$$(s_0, s_1, s_2, \dots, s_{N-1}) = \underbrace{(s_0, 0, 0, \dots, 0)}_{\text{DFT has flat magnitude}} + \dots + (0, 0, \dots, 0, s_{N-1})$$





Proof - continued



$(\mathbf{0}, s_1, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0})$: 1 left-cyclic shift of $(s_1, \mathbf{0}, \mathbf{0}, \dots, \mathbf{0})$

Cyclic shift in time domain does not affect on
the magnitude in the frequency domain

Proof - continued

$(0, s_1, 0, 0, \dots, 0)$: 1-cyclic shift of $\underline{(s_1, 0, 0, \dots, 0)}$

Cyclic shift in time domain does not affect on magnitude at the frequency domain

$$\begin{array}{c}
 \left(\begin{array}{cccc} \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \end{array} \right) \otimes \left(\begin{array}{cccc} (\omega^1)^0 & 0 & \cdots & 0 \\ (\omega^1)^1 & 0 & \cdots & 0 \\ (\omega^1)^2 & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ (\omega^1)^{m-1} & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ (\omega^1)^{m(N-1)} & 0 & \cdots & 0 \\ (\omega^1)^{m(N-1)+1} & 0 & \cdots & 0 \\ (\omega^1)^{m(N-1)+2} & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ (\omega^1)^{mN-1} & 0 & \cdots & 0 \end{array} \right)
 \end{array}$$

$$(\omega^1)^x = \left(e^{-\frac{j2\pi}{mN}} \right)^x = \left(e^{-\frac{j2\pi}{mN^2}} \right)^{Nx}$$

Proof - continued

$(0, s_1, 0, 0, \dots, 0)$: 1-cyclic shift of $\underline{(s_1, 0, 0, \dots, 0)}$

Cyclic shift in time domain does not affect on magnitude at the frequency domain

$$\left(\begin{array}{cccc} \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \end{array} \right)$$



$$\left(\begin{array}{cccc} \left(e^{-\frac{j2\pi}{mN^2}}\right)^0 & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{N-1} \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^N & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{2N-1} \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^{2N} & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{3N-1} \\ \vdots & \vdots & \vdots \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^{(m-1)N} & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{mN-1} \\ \vdots & \vdots & \vdots \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^{m(N-1)N} & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{m(N-1)N+N-1} \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^{[m(N-1)+1]N} & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{[m(N-1)+1]N+N-1} \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^{[m(N-1)+2]N} & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{[m(N-1)+2]N+N-1} \\ \vdots & \vdots & \vdots \\ \left(e^{-\frac{j2\pi}{mN^2}}\right)^{(mN-1)N} & \cdots & \left(e^{-\frac{j2\pi}{mN^2}}\right)^{(mN-1)N+N-1} \end{array} \right)$$

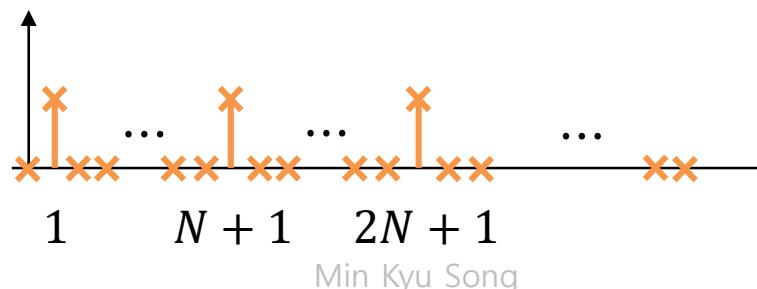
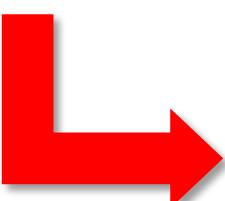
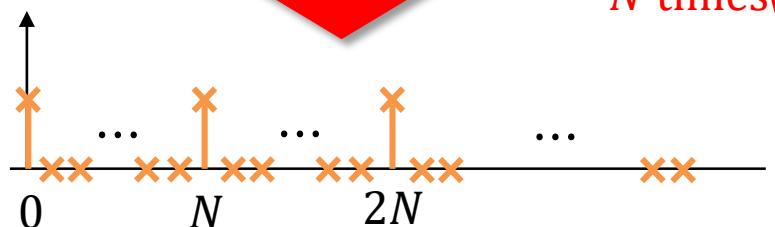
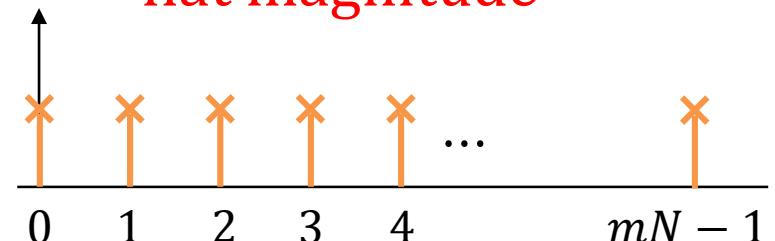
Linear phase shift

Proof - continued

$(0, s_1, 0, 0, \dots, 0)$: 1-cyclic shift of $\underline{(s_1, 0, 0, \dots, 0)}$

Cyclic shift in time domain does not affect on magnitude at the frequency domain

DFT has flat magnitude



$$\begin{pmatrix} \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(0) & 0 & \cdots & 0 \\ \alpha(1) & 0 & \cdots & 0 \\ \alpha(2) & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ \alpha(m-1) & 0 & \cdots & 0 \end{pmatrix}$$



$$\begin{pmatrix} (\omega^1)^0 & 0 & \cdots & 0 \\ (\omega^1)^1 & 0 & \cdots & 0 \\ (\omega^1)^2 & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ (\omega^1)^{m-1} & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ (\omega^1)^{m(N-1)} & 0 & \cdots & 0 \\ (\omega^1)^{m(N-1)+1} & 0 & \cdots & 0 \\ (\omega^1)^{m(N-1)+2} & 0 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ (\omega^1)^{mN-1} & 0 & \cdots & 0 \end{pmatrix}$$

1-cyclic shift at the frequency domain

Proof - continued

Time domain

$$(s_0, 0, 0, \dots, 0)$$

+

$$(0, s_1, 0, 0, \dots, 0)$$

+

\vdots

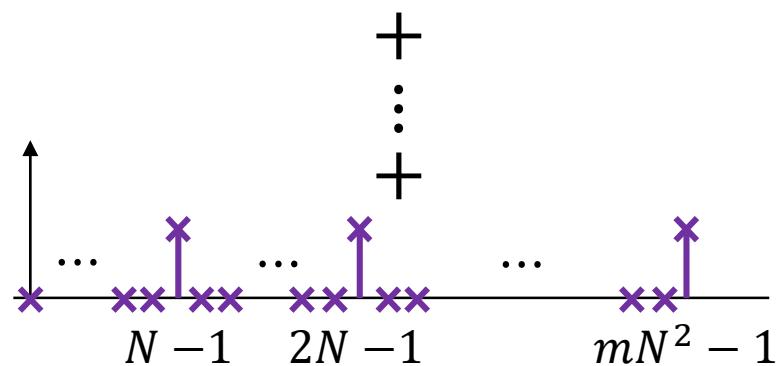
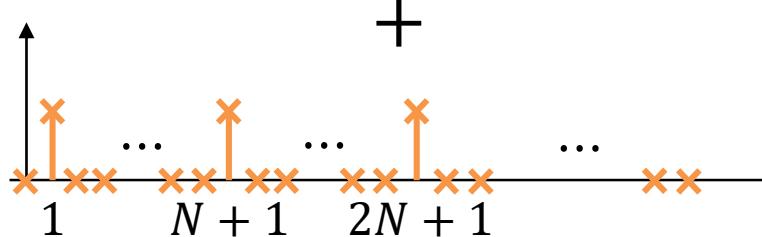
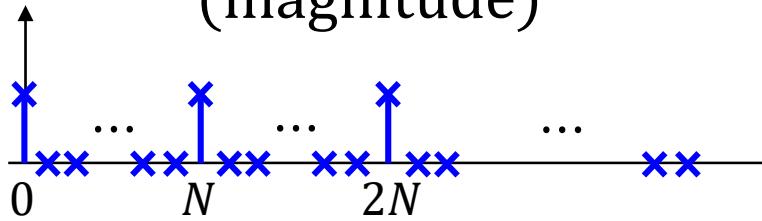
$$(0, 0, \dots, 0, s_{N-1})$$

||

$$(s_0, s_1, s_2, \dots, s_{N-1})$$



Frequency domain
(magnitude)



?

Time domain

$$(s_0, 0, 0, \dots, 0)$$

+

$$(0, s_1, 0, 0, \dots, 0)$$

+

\vdots

+

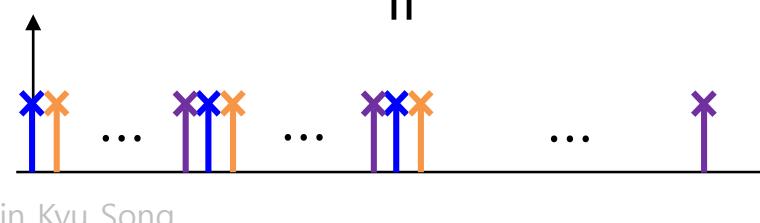
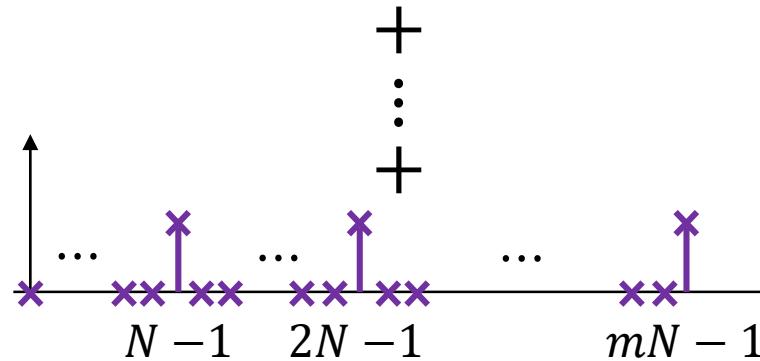
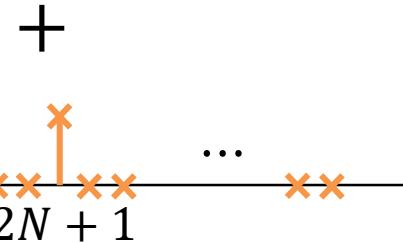
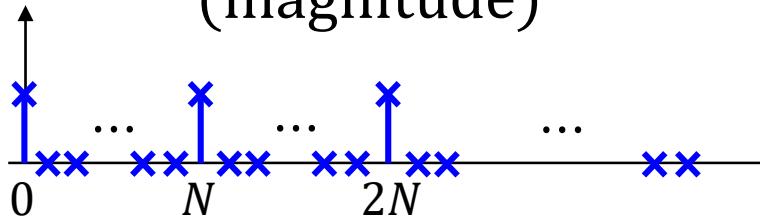
$$(0, 0, \dots, 0, s_{N-1})$$

||

$$(s_0, s_1, s_2, \dots, s_{N-1})$$

Proof - done

Frequency domain
(magnitude)



Examples

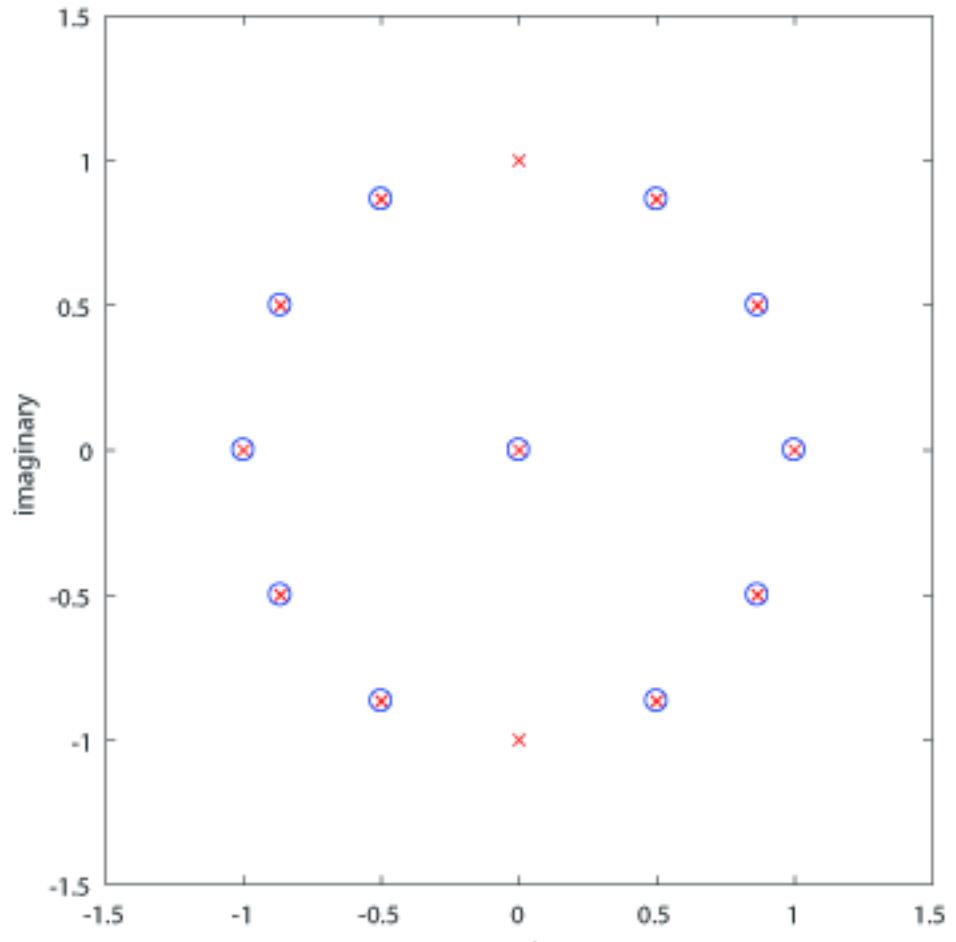
A perfect sequence

$$\{0, -1, 1, 0, 1, 1\} \quad N = 2 \quad \text{all-one}$$

**Generalized
Milewski
Construction**

$$s = \{0, 0, -1, -\omega, 1, \omega^2, 0, 0, 1, \omega^4, 1, \omega^5, 0, 0, -1, -\omega^7, 1, \omega^8, 0, 0, 1, \omega^{10}, 1, \omega^{11}\}$$

$$\tilde{\omega} = e^{-\frac{j2\pi}{12}}$$



Constellation of s : 12-PSK+

Examples (cont')

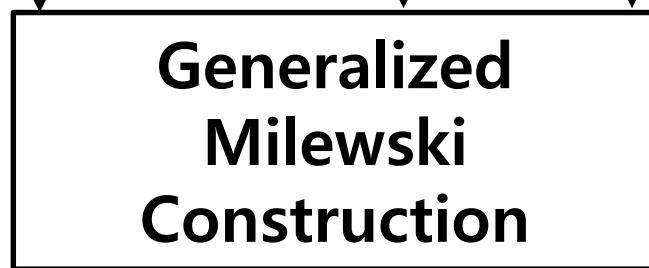
A perfect sequence

$$\{3, -2, 3, -2, -2, 3, -2, -7, -2, -2\}$$

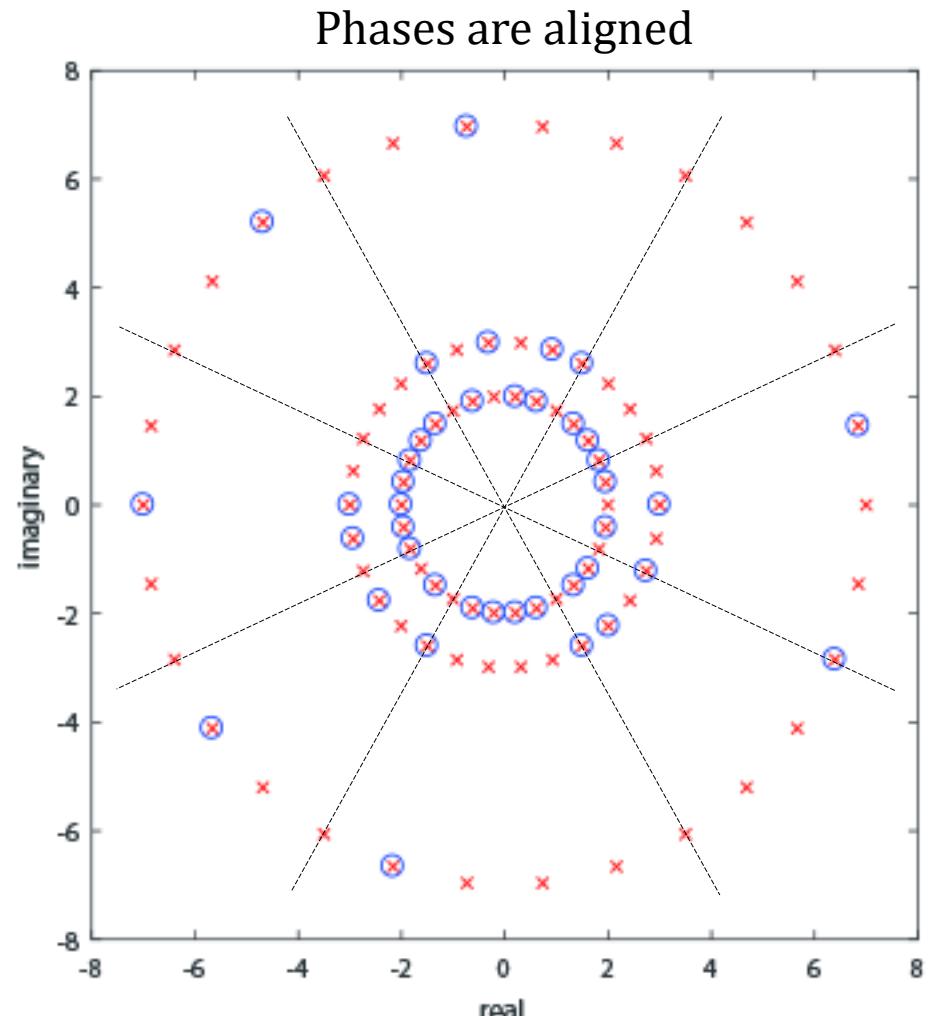
of period 9

$$N = 3$$

all-one



s is a perfect sequence of period 90



How many different phases?

※ ω is an mN -th primitive root of unity

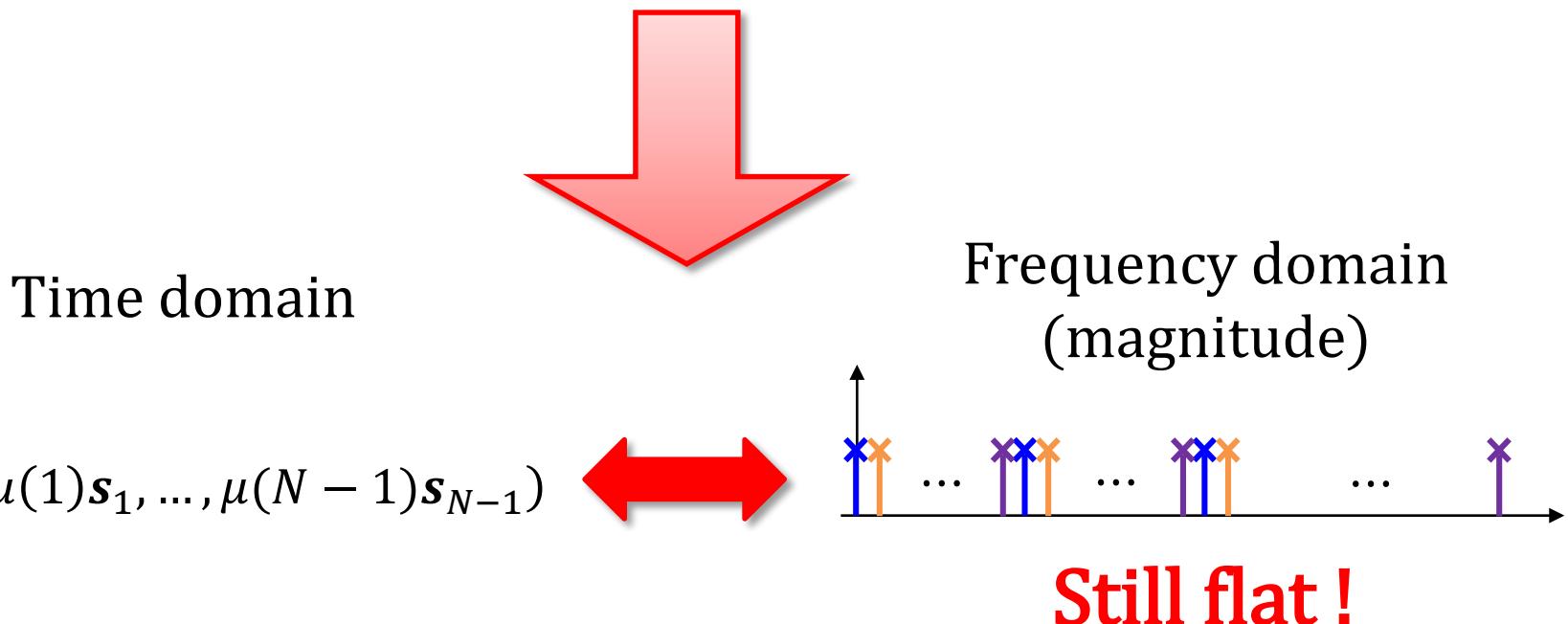
$\alpha(0)$	$\times (\omega^0)^0$	$\alpha(0)$	$\times (\omega^1)^0$...	$\alpha(0)$	$\times (\omega^{N-1})^0$
$\alpha(1)$	$\times (\omega^0)^1$	$\alpha(1)$	$\times (\omega^1)^1$...	$\alpha(1)$	$\times (\omega^{N-1})^1$
$\alpha(2)$	$\times (\omega^0)^2$	$\alpha(2)$	$\times (\omega^1)^2$...	$\alpha(2)$	$\times (\omega^{N-1})^2$
\vdots		\vdots		\ddots		\vdots
$\alpha(m - 1) \times (\omega^0)^{m-1}$		$\alpha(m - 1) \times (\omega^1)^{m-1}$...	$\alpha(m - 1) \times (\omega^{N-1})^{m-1}$	
\vdots		\vdots		\ddots		\vdots
$\alpha(0)$	$\times (\omega^0)^{m(N-1)}$	$\alpha(0)$	$\times (\omega^1)^{m(N-1)}$...	$\alpha(0)$	$\times (\omega^{N-1})^{m(N-1)}$
$\alpha(1)$	$\times (\omega^0)^{m(N-1)+1}$	$\alpha(1)$	$\times (\omega^1)^{m(N-1)+1}$...	$\alpha(1)$	$\times (\omega^{N-1})^{m(N-1)+1}$
$\alpha(2)$	$\times (\omega^0)^{m(N-1)+2}$	$\alpha(2)$	$\times (\omega^1)^{m(N-1)+2}$...	$\alpha(2)$	$\times (\omega^{N-1})^{m(N-1)+2}$
\vdots		\vdots		\ddots		\vdots
$\alpha(m - 1) \times (\omega^0)^{mN-1}$		$\alpha(m - 1) \times (\omega^1)^{mN-1}$...	$\alpha(m - 1) \times (\omega^{N-1})^{mN-1}$	

mN different phases required, in general

For arbitrary chosen μ

Multiplying a constant $\mu(r)$ with $|\mu(r)| = 1$
to the r -th column
does not affect on the magnitudes of the DFT of

$$(\mathbf{0}, \mathbf{0}, \dots, \mu(r)s_r, \mathbf{0}, \dots, \mathbf{0}) = \mu(r)(\mathbf{0}, \mathbf{0}, \dots, s_r, \mathbf{0}, \dots, \mathbf{0})$$



A comparison

$$\gcd(L_1, L_2) = 1$$

perfect sequence

$$x = \{x(n)\}_{n=0}^{L_1-1}$$

perfect sequence

$$y = \{y(n)\}_{n=0}^{L_2-1}$$

Direct product construction

$$\text{Output perfect sequence } z = \{z(n)\}_{n=0}^{L_1 L_2 - 1}$$

$$= \{x(n)y(n)\}_{n=0}^{L_1 L_2 - 1}$$

- Synthesize a long perfect sequence
- Need **two** short **perfect sequences** of **relatively prime period**
- The **constellation** depends on **input** perfect sequences

Perfect sequence

$$\alpha = \{\alpha(n)\}_{n=0}^{m-1}$$

A positive integer

$$N$$

Any polyphase sequence of length N

$$\mu$$

Generalized Milewski Construction

$$\text{Output perfect sequence } s = \{s(n)\}_{n=0}^{mN^2 - 1}$$

$$= \{\mu(r)\alpha(q)\omega^{qr}\}_{n=0}^{mN^2 - 1}$$

- Synthesize a long perfect sequence
- Need a **single** short **perfect sequence**
- **PSK+** or **APSK+** constellations

Concluding remarks

- The Milewski construction is generalized **in period and constellation**
 - flexible period, various input sequences
 - perfect sequences over PSK, PSK+, APSK+
 - Example: Input perfect over ASK+ → Output perfect over APSK+

- Some interesting questions for future works:
 - **Possible to reduce the number of phase?**
 - At least mN different phases, in general.
 - related to constructing perfect sequences over QAM constellation
 - **For practical APSK, how to tilt points?**
 - Two points on circles of different radius have different phase offset to maximize distance, in practice.

